Analog and Digital Circuits Test 4

Number of Questions: 35

Directions for questions 1 to 35: Select the correct alternative from the given choices.

- 1. The output impedance of a BJT under common collector configuration is
 - (A) high (B) medium
 - (C) low (D) very high
- 2. Body effect in MOSFETS results in
 - (A) Increase in the value of output resistance
 - (B) Increase in the value of transconductance
 - (C) change in the value of threshold
 - (D) decrease in the value of transconductance
- 3. The modified work function of an *n*-channel MOSFET is -0.75V. If the interface charge is 2×10^{-4} C/m² and the oxide capacitance is $250\mu F/m^2$, the flat band voltage is
 - (B) 1.25 V (A) 1.4 V
 - (C) 1.75 V (D) 0.2 V
- 4. A circuit using the BJT is shown in the below figure, the value of β is



(11)	250	(D)	175
(C)	176	(D)	249

5. For the circuit shown in the below figure, by assuming $\beta = 150$ and $V_{BE} = 0.7$ V, the best approximation for the collector current I_c in the active region is



- 6. The purpose of emitter capacitor across R_{F} is to
 - (A) reduce noise in the amplifier
 - (B) protect the transistor
 - (C) avoid voltage gain drop
 - (D) provide biasing
- 7. What are the effects of cascode amplifier stages?
 - (1) Low input impedance
 - High input impedance (2)
 - (3) Wideband amplifier
 - (4) Narrow band amplifier
 - (A) (1) and (3) only
 - (B) (2) and (3) only (C) (1) and (4) only (D) (2) and (4) only
- 8.



The value of β is

 $\left(\frac{V_0}{V}\right)$ is 9. For the circuit shown in figure, the voltage gain



Section Marks: 90

- **10.** Which one of the following oscillators is well suited for the generation of wide range audio frequency waves?
 - (A) Hartley oscillator
 - (B) Wein bridge oscillator
 - (C) Colpitt's oscillator
 - (D) crystal oscillator
- **11.** If $a = (b + c) (b^1 + c^1)$, then the value of *b* is

(A)
$$(a^{1} + c)(a + c^{1})$$
 (B) $a^{1}c^{1} + ac^{1}$
(C) $a^{1}c^{1} + a^{1}c$ (D) $(a^{1} + c^{1})(a + c^{1})$

- **12.** The max term expression of a four variable even function is?
 - (A) $\pi M(0, 2, 4, 6, 8, 10, 12, 14)$
 - (B) $\pi M(1, 3, 5, 7, 9, 11, 13, 15)$
 - (C) $\pi M(0, 3, 5, 6, 9, 10, 12, 15)$
 - (D) $\pi M(1, 2, 4, 7, 8, 11, 13, 14)$
- 13. Find which of the following is not a minimum sum of the products expression for $f(w, x, y, z) = \Sigma m(0, 3, 5, 7, 8, 9, 10, 12, 13) + d(1, 6, 11, 14)$
 - (A) $\overline{wz} + \overline{xy} + w\overline{x} + \overline{yz}$ (B) $w\overline{x} + w\overline{y} + \overline{xy} + \overline{wz}$
 - (C) $\overline{xy} + \overline{wz} + w\overline{z} + \overline{yz}$ (D) $w\overline{z} + \overline{xy} + w\overline{y} + \overline{wz}$
- 14. The input clock frequency is 10 kHz, then the frequency of *Y* is (initially Q = 0)



(A)	5 kHz	(B)	10 kHz
(C)	2.5 kHz	(D)	0 Hz

- **15.** If $(3.5)_{base 6} + (2.3)_{base 6} = (X)_{base 6}$ then the value of X is (A) 5.8 (B) 10.2 (C) 6.2 (D) 5.6
- 16.



Consider diodes are ideal, If $V_1 = -2V$, find the value of V_0 .



- Find the voltage across capacitor V_{c} .
- (A) -0.023 V (B) 9.91 V
- (C) 19.915 V (D) -0.085 V
- **18.** The data sheet for an E-MOSFET gives $I_{D(on)} = 4$ mA at $V_{GS} = 10$ V and $V_{GS(th)} = 2$ V. Then find the value of *K*. (A) 62.5 μ A/V² (B) 54.75 μ A/V²
 - (A) $62.5 \,\mu\text{A/V}$ (B) $54.75 \,\mu\text{A/V}$ (C) $6.25 \,\text{mA/V}^2$ (D) $5.47 \,\text{mA/V}^2$
- **19.** For the transistor circuit shown in figure $I_R = 35 \ \mu A$ and $V_{RF} = 0.7 \text{V}$

7.5V

$$R_{c} \ge 2k\Omega$$

 $\beta = 99$
 R_{B}
 $R_{E} \ge 1.25k\Omega$

The value of R_{B} would be

(A)	55 kΩ	(B)	5.23 kΩ
(C)	69.28 kΩ	(D)	$7.28 \text{ k}\Omega$

20. A 7.5V zener diode is used in the circuit shown in figure, and the load current is to vary from 15 to 80mA. Find the value of series resistance '*R*' to maintain a voltage of 7.5V across the load, consider the $I_{Z(min)} = 8$ mA.





(B) -40 V

The resulting output V_0 is (A) zero

- (C) -10 V (D) -8 V
- **22.** The tuned collector oscillator circuit used in the local oscillator of a radio receiver makes use of an *LC* tuned circuit with $L_1 = 50\mu$ H, and $C_1 = 250$ pF. Then the frequency of oscillators is

(A)	14.23 KHz	(B)	1.423 MHz
(C)	284 MHz	(D)	1.5 MHz

23. In a current amplifier, with A = 1000 and $\beta = 0.25$, the input resistance is 100 Ω before negative feedback is applied. After negative feedback is applied it's input resistance will be

(A)	0.4Ω	(B)	251Ω
(C)	150Ω	(D)	100Ω

24. The Schmitt trigger circuit is shown in the below figure. If $V_{sat} = \pm 12$ V, the upper threshold voltage would be



25. In the rectifier circuit shown in figure. The minimum peak – inverse – voltage (PIV) rating of the diode is



Consider diode D is ideal.

(A)
$$10 V$$

(C) $10\sqrt{2} V$

26.



(B) $20\sqrt{2}$ V

(D) 20 V

The Function F(a, b, c) is

(A)	$a^1 + b^1 + c$	(B)	$(abc)^1$
(C)	$(a^1 + b^1)c^1$	(D)	$(a+b)^{1}+c^{1}$

27. If the following circuit is converted to all-NAND network, then how many number of NAND gates are required? (inverted inputs are available)



28. A 3 bit Digital to Analog converter is modified as follows, to make it 6 bit Digital to analog converter. So what will be the value of resistor '*r*' for proper operation?



29. The switches S_i will be connected to $V_R = 1$ V for logic 1, and will be connected to 0V for logic 0. Find the output V_o for input $S_3S_2S_1S_0 = 0101$

A, Byte1

Output

А

А

Port

0

1

2

3

0

1

2

3

(B) $\overline{A}\overline{C} + B\overline{C}$

(D) $\overline{A}C + B\overline{C}$

(B) $\overline{A} + C$

(D) $A + \overline{C}$

MVI

ORA

XRA

OUT

HLT

 2×4

Decoder

EN

EN

 2×4 Decoder

Β.

B

B,

B

JM



				ANSV	VER KEYS				
1. C	2. C	3. A	4. D	5. D	6. C	7. B	8. A	9. C	10. B
11. B	12. D	13. A	14. B	15. B	16. C	17. D	18. A	19. C	20. A
21. A	22. B	23. A	24. A	25. B	26. B	27. A	28. C	29. A	30. D
31. D	32. C	33. D	34. B	35. C					

INTS	AND	Expla	NATION	S

Choice (C)

1. For CC amplifier $R_{in} = high$ $R_o^{m} = 1$ ow

2. Choice (C)

3. We know that flat band voltage
$$V = \frac{q}{C} (1 - W_c)$$
 volts
= $\frac{2 \times 10^{-4}}{250 \times 10^{-6}} (1 + 0.75) = 1.4$ volts Choice (A)

 α

4.
$$I_E = \frac{15 - 2.5}{2.5 \text{k}\Omega} = 5 \text{mA}$$

 $I_B = \frac{2}{100} \text{mA} = 20 \mu \text{A}$
 $(1+\beta) = \frac{I_E}{I_B} = 250$
 $\therefore \beta = 249$

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- 5. Applying *KVL* to the input loop $1.2 - I_B R_B - V_{BE} - R_E I_E = 0$ $1.2 - 0.7 = I_B [10 \times 10^3 + 0.5 \times 10^3 \times 151]$ $I_B = 5.8479 \mu A$ $I_C = \beta I_B$ $= 150 \times 5.8479 \mu A$ = 0.877 m A Choice (D)
- 6. For *D*, *C* analysis R_E advantage it improves the stability. But *A*. *C* analysis R_E decrease the gain.
 - \therefore Capacitor across R_E it eliminates the R_E effect
 - $\therefore \text{ Short circuit parallel to } R_E \qquad \qquad \text{Choice (C)}$
- 7. (1) Cascode amplifier (CE CB) gives the high input impedance
 - (2) It is providing high frequency performance of CB
 - (3) It is also used as wide band amplifier Choice (B)
- **8.** From the given circuit

$$I_{C} = \frac{2}{500} = 4 \text{ mA}$$

$$I_{B} = \frac{4.5 - 2}{15} \text{ mA} = 0.1666 \text{ mA}$$

$$I_{C} = \beta I_{B}$$

$$\beta = \frac{I_{C}}{I_{B}} = 24$$
Choice (A)

9. Applying virtual GND concept and redrawing the given circuit.



- **10.** Audio frequency oscillators (<20 KHz)
 - (1) RC phase shift
 - (2) Wein bridge
 - RF oscillators (> 20 KHz)

(2) Crystal oscillator (3) Hartly oscillator (4) Clapp oscillatoretc Choice (B) 11. From given $a = (b + c)(b^1 + c^1)$ $= bc^1 + b^1 c = b \oplus c$ EXOR with c both sides $a \oplus c = b \oplus c \oplus c = b \oplus 0 = b$ So $b = a \oplus c = a^1 c + ac^1$ or $(a + c)(a^1 + c^1)$. Choice (B)

(1) Colpitt's oscillator

- **12.** Even function is a Boolean function, and it will be equal to 1, if the input variables have an even number of 0's.
 - (odd function is a Boolean function for which output is 1, for input combinations with odd number of 1's) So for 4 variables, even number of zeroes occur for input combinations 0000, 0011, 0101, 0110, 1001, 1010, 1100, 1111 For these combinations, even function output is 1. So remaining terms are max terms.

So
$$f_{\text{even}} = \Pi M(1, 2, 4, 7, 8, 11, 13, 14)$$
 Choice (D)

-1	-	
	- 1	
-	~	ĺ

yz wx	z 100	01	11	10
00	1	X	1	
01		1	1	Х
11	1	1		х
10	(1	1	х	1
	1			

x y, wz is the essential prime implicants, remaining prime implicants can be used to cover all the min terms.

yz wx	z 100		01		11	10
00	1		X)	1	
01			1		1	х
11	1		1			х
10	(1		1		Х	1
	4	_		Ľ		

wz, xy, wx, yz will not implement the *k*-map as in the above *k*-map min term wxyz = 1100 is not covered. Choice (A)

14. When Q = 0, initially, $\overline{Q} = 1$

So J = K = 1, by applying first clk pulse $Q_{n+1} = 1$, Then $\overline{Q}_{n+1} = 0$. So J = 1, K = 0, after clock pulse $Q_{n+2} = 1$, ... and for next clock pulses it will remain in same state, so the frequency of Q = 0Hz (Q = 1 always) but Q = 1, so

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the clk pulse ANDed with Q = 1 and Y is same as clk pulse so frequency of *Y* is 10 kHz. Choice (B) **15.** 3.5 + 2.3 ? 21. 5 + 3 = 8 $(12)_6$ so result is 2 and carry is 1. 1 3.5 +2.3?.2 1 + 3 + 2 = 6 $(10)_6$ so result is 0 and carry is 1 3.5 + 2.3 10.2 Choice (B) **16.** From the given circuit $D_1 \rightarrow ON, D_2 \rightarrow OFF$.:. It becomes 1.5 mA $1 k\Omega$ 15 V -2V WW -> ↓ V₀

$$V_o = 1.5 - 2$$

= -0.5 volts Choice (C)

17. We know
$$I_E = I_B + I_C$$

 $I_E = \frac{12 - 2}{5} \text{mA} = 2\text{mA}$

$$I_{B} = \frac{1.7}{100} \text{ mA} = 0.017 \text{ mA}$$

$$I_{C} = 1.983 \text{ mA}$$

$$I_{C} = \frac{V_{C} + 10}{5} \text{ mA}$$

$$V_{C} + 10 = 1.983 \times 5$$

$$V_{C} = 9.915 - 10 = -0.085 \text{ volts.}$$
Choice (D)

18. $I_D = k[V_{GS} - V_{GS(th)}]^2$ Amp K =

$$K = \frac{1}{\left[V_{GS(on)} - V_{GS(th)}\right]^{2}}$$

= $\frac{4 \times 10^{-3}}{\left[10 - 2\right]^{2}} = 62.5 \times 10^{-6} \frac{A}{V^{2}}$ Choice (A)

19. Applying KVL to input loop

$$7.5 - R_B \cdot I_B - V_{BE} - R_E \cdot I_E = 0$$

 $I_E = (1 + \beta) \cdot I_B$
 $= 100 \times 35 \times 10^{-6}$
 $= 3.5 \text{ mA}$
 $7.5 - 0.7 - 4.375 = R_B \cdot I_B$
 $R_B = \frac{2.425}{35} \times 10^6 = 69.28 \text{ k}\Omega$ Choice (C)]

20.
$$I = I_Z + I_L$$

 $I_{R(max)} = I_{Z(min)} + I_{L(max)} = (8 + 80) \text{ mA} = 88 \text{ mA}$
 $R = \frac{15 - 7.5}{88} \text{k}\Omega = 85.22\Omega$ Choice (A)



The o/p voltage of 1st stage is
$$V_{01}$$

 $V_{01} = \frac{-8}{2} \times 2 = -8$ volts

Redrawing the given circuit $V_0 = -5 \times (-8) - 10 \times 4$ = 40 - 40 = 0V

22. Frequency of oscillation ,
$$f = \frac{1}{2\pi\sqrt{L_1C_1}}$$

$$f = \frac{1}{2\pi\sqrt{50 \times 10^{-6} \times 250 \times 10^{-12}}} = \frac{1 \times 10^{+8}}{2\pi\sqrt{125}}$$

$$2\pi\sqrt{30 \times 10^{-1} \times 250 \times 10^{-1}} = 2\pi\sqrt{125}$$

$$= 1.426 \text{ MHz} \qquad \text{Choice (B)}$$
23. From the given data
$$A = 1000; \beta = 0.25$$

$$R_i = 80\Omega$$

$$1 + A\beta = 1 + 1000 \times 0.25 = 251$$
Current amplifier
$$i/p \Rightarrow \text{current source}$$

$$i.e, \text{ it is a current shunt amplifier}$$

$$\therefore R_i \Rightarrow \text{ decreases}$$

$$R_0 \Rightarrow \text{ increases}$$

$$R_i = 100$$

$$\therefore \quad R_{if} = \frac{R_i}{1 + A\beta} = \frac{100}{251} \approx 0.4\Omega \qquad \text{Choice (A)}$$

V₀

ξ 45KΩ

↓V_{υτ}

0.2V

≶ 5ΚΩ

24. Max. output
$$V_0 = \pm V_{sat} = \pm 12V$$

If $V_{in} = V_{UT}$ redrawing the given circuit.
 $V_+ = V_- = V_{UT}$

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Applying nodal analysis at node V_{UT}

$$\frac{V_{UT} - V_0}{45K} + \frac{V_{UT} - 0.2}{5K} = 0$$

$$V_{UT} - V_0 + 9V_{UT} - 1.8 = 0$$

$$10V_{UT} = V_0 + 1.8$$
But $V_0 = +V_{sat} = +12V$

$$V_{UT} = 1.38$$
 Volts Choicer (A)

- **25.** During +*Ve* half cycle $D \rightarrow ON$
 - \therefore Voltage across capacitor $V_c = 10 V_{\text{rms}}$.



 $\begin{array}{ll} \therefore & V_{_{C}} = 10 \mathrm{V}_{_{\mathrm{rms}}}.\\ & \mathrm{During} - Ve \text{ half cycle } D \rightarrow \mathrm{OFF} \end{array}$



PIV = $V_d = (10 + 10)V_{\text{rms}} = 20\sqrt{2}$ volts. Choice (B)

- **26.** Output of 1^{st} MUX $Y_1 = 1.a^1 + b^1.a = a^1 + b^1$ Output of 2^{nd} MUX $Y_2 = 1.a + 1.a^1 = 1$ Output of 3^{rd} MUX $F(a, b, c) = Y_1.c + Y_2.c^1$ $= (a^1 + b^1)c + 1.c^1$ $= a^1 + b^1 + c^1$ Choice (B)
- 27. Given function has to be in SOP form, so it can be implemented by AND-OR gates, same as NAND-NAND gates

$$f(a, b, c, d, e) = (a + b^{1})c + cd^{1} + (d^{1} + e)$$

= $ac + b^{1}c + cd^{1} + d^{1} + e$

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$$= ac + b^1 c + d^1 + e$$

28. The $S_2 S_1 S_0$ are lower order bits, $S_5 S_4 S_3$ are higher order bits so the current due to bits $S_2 S_1 S_0$ should be one-eighth of the current due to $S_5 S_4 S_3$.

The resistor r has been inserted to provide the attenuation.

Choice (A)

Such attenuation, is possible when r = 4R. Choice (C)

29. Given circuit is a DAC. (R - 2 R type)So the voltage at non inverting terminal is

$$V_{+} = \frac{V_{R}}{2^{n}} \left(S_{0} \times 2^{0} + S_{1} \times 2^{1} + S_{2} \times 2^{2} + S_{3} \times 2^{3} \right)$$

$$= \frac{1}{16} \times (1+0+4+0) = \frac{5}{16}$$

$$V_0 = \left(1 + \frac{R}{2R}\right) \frac{5}{16} = \frac{3}{2} \times \frac{5}{16} = \frac{1}{2} \times \frac{15}{16}$$

$$= 0.46875 \text{ V} \qquad \text{Choice (A)}$$

- **30.** Address of last memory location Address of first Memory location = (Number of address locations – 1) Here given RAM number of address locations = 2048 $(2048)_{10} = 0800H$ 0800H - 1 = 07FFHX - 0900H = 07FFHX = 07FF + 0900H = 10FFH Choice (D)
- 31. Noise Margin

 $\Delta 1 = -0.76 - (-1.1) = 0.34V$ $\Delta 0 = -1.25 - (-1.58) = 0.33V$ Noise margin is min ($\Delta 0$, $\Delta 1$) = 0.33V. **Fanout**:- $N_1 = \frac{I_{OH}}{I_{H}} = \frac{3\text{mA}}{107\mu\text{A}} = 28$

$$N_0 = \frac{I_{OL}}{I_{IL}} = \frac{3.7 \text{mA}}{137 \mu \text{A}} = 27$$

Fanont = min $(N_1, N_0) = 27$

Choice (D)

- **32.** As the above programs ends at memory location 400DH, the address of next memory location is 400EH, which will be stored in Program Counter at the end of the program.
 - L X I SP, 5000H \rightarrow load SP = 5000H L X I H, 5050H \rightarrow load HL = 5050H SPHL \rightarrow move HL to SP, so SP = 5050H PUSH H \rightarrow SP decremented by 2 (SP - 2) PUSH B \rightarrow SP decremented by 2 (SP - 4) CALL 4050H \rightarrow after call instruction, again returned to next instruction, so no charge for SP. POP H \rightarrow increment SP by 2 (SP - 4 + 2) = SP - 2 HLT - stop So contents of stack pointer are 5050-2 = 504EH. Choice (C)
- 33. MVI A, Byte1 → A = Byte1, copy number to A.
 ORA A make A + A = A, contents of Accumulator remain same but CY = 0, and other flags will change as per number in Accumulator.
 JM output Jump on minus to output port if S = 1 go

to output. $XR \land \land$ else $XOR \land$ with $\land \land \oplus \land = 00U$

XRA A – else XOR A with A, A
$$\oplus$$
 A = 00H
A = 00H

OUT port \rightarrow send the contents of Acc to output port HLT – stop.

So if there is any negative number it will be the output or zero is the output. Choice (D)

34. Given is a 3×8 Decoder with two 2×4 Decoders, and a NOT gate, LSB are connected to inputs of Decoder

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 $(B_1 B_0 = BC)$, MSB (A) is connected to check which decoder has to be selected by using enable input, When A = 0, first Decoder will be selected, when A = 1, second decoder will be selected.

 f_1 is connected through NOR gate

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outputs of Decoder (here active High outputs) are min terms, min terms connected to NOR gate.

 $f_1 = \pi M(1, 3, 4, 5, 7) = \Sigma m(0, 2, 6)$ because in second decoder A = 1, BC vary as per inputs applied so 4, 5, 6, 7 are output of second decoder (min terms)



$$f_1 = AC + BC$$

35.
$$f_2$$
 is connected to *OR* gate, so sum of min terms $f_2 = \Sigma m(0, 2, 3, 4, 6, 7) = \pi M(1, 5)$

$$f_{2} = (B + \overline{C})$$

$$f_{2} \xrightarrow{BC} 00 \quad 01 \quad 11 \quad 10$$

$$0 \quad 0 \quad 0 \quad 1$$

$$f_{2} = (B + \overline{C})$$

Choice (C)