## **COMMUNICATIONS TEST I**

## Number of Questions: 35

# *Directions for questions 1 to 35:* Select the correct alternative from the given choices.

- In a digital communication system FSK signal used to transmit 0 and 1, with the help of two sine waves having frequency 25KHz & 50KHz respectively. The minimum bit duration for these waveforms to be orthogonal is
  - (A) .04ms (B) .08ms
  - (C) 0.12ms (D) 0.06ms
- 2. For the modulated carrier  $x(t) = 400 \cos(\omega_c + 100\pi)t + 2000\cos\omega_c t + 400 \cos(\omega_c 100\pi)t$  where  $2000 \cos\omega_c t$  is the unmodulated carrier, the complex envelope is \_\_\_\_\_.
  - (A) 400  $\cos(100\pi t)$
  - (B)  $2000 + 400 \cos(100\pi t)$
  - (C)  $2000 800 \cos(100\pi t)$
  - (D)  $2000 + 800 \cos(100\pi t)$
- 3. Match List I (operations) with List II (functions) and select the correct answer using the codes given below the lists:

List-I	List-II
p. companding	1. Improving image rejection ration
q. pre emphasis	2. muting the receiver
r. squelch	3. non-uniform quantization
s.Double conversion	4. Boosting of higher modulating frequencies at the transmitter

**Codes:** 

(A) <i>p</i> −1	q-3	r - 1	s –2
(B) $p-3$	q-2	r-4	s-1
(C) <i>p</i> -2	q-4	r - 1	s-3
(D) $p - 3$	q-4	r-2	s-1

4. Match the following

	(i)	PSK		1	Matched	filter	
	(ii)	FM		2	PLL		
	(iii)	AM		3	Synchron	ous detector	
	(iv)	DSB		4	Envelope Detector		
(A)	(i) –	4	(ii) – 2		(iii) –3	(iv) – 1	
(B)	(i) –	1	(ii) – 2		(iii) – 4	(iv) – 3	
(C)	(i) –	2	(ii) – 1		(iii) – 4	(iv) – 3	
(D)	(i) –	4	(ii) – 2		(iii) – 1	(iv) – 3	

5. The message signal  $m(t) = \operatorname{sinc}^2(4000t)$  modulates a carrier  $\cos(2\pi fct)$  and generate a SSB-SC signal. The *B*. *W* of SSB – SC is

5	KHz
	5

- (C) 2 KHz (D) 1 KHz
- 6. A binary sequence of bits 1101100 is differentially encoded and phase shift keyed and producing a

transmitted phase for the first bit 1 as  $\pi$ . The transmitted phase for the following bits (101100) are \_\_\_\_\_.

- (A)  $(\pi, 0, 0, 0, \pi, 0)$  (B)  $(\pi, \pi, 0 \ 0 \ 0 \ \pi)$
- (C)  $(0, \pi, \pi, \pi, 0, \pi)$  (D)  $(0, 0, \pi, \pi, \pi, 0)$
- 7. Consider following statements about PCM(1) Bandwidth depends on the sampling rate
  - (2) Cross talk occurs in PCM
  - (3) Band width depends on the quantization error
  - (4) Quantization error depends upon the step size. True statements are \_\_\_\_\_
  - (A) 1 and 2 (B) 2 and 3
  - (C) 3 and 4 (D) 4 and 1
- 8. Match the list -I & List II

	List-I		List-II		
Р	AMI	1	Pusle stuffing		
Q	Asynchronous TDM	2	Voice channels		
R	T4 carrier system	3	Clock recovery		
S	RZ code	4	Scrambling		
(A) $P-3, Q-2, R-4, S-1$					
(B)	P-4, Q-1, R-2, S-	- 3			

- (C) P-4, Q-1, R-3, S-2(D) P-3, Q-2, R-1, S-4
- **9.** In a PCM system with 4 quantization levels, the maximum quantization noise power will be \_\_\_\_\_
  - (A)  $\left(\frac{1}{8}\right)^{th}$  of total (amplitude range)<sup>2</sup>
  - (B)  $\left(\frac{1}{192}\right)^{th}$  of total (amplitude range)<sup>2</sup>
  - (C)  $\left(\frac{1}{8}\right)^m$  of total amplitude range
  - (D)  $\left(\frac{1}{192}\right)^{th}$  of total amplitude range
- In a super heterodyne receiver double spotting phenomena occurs at 2500 KHz. If intermediate frequency is 450KHz. Then the receiver should be tuned at
   (A) 2050 KHz
   (B) 2400 KHz
  - (A) 2050 KHz
     (B) 3400 KHz
     (C) 1600 KHz
     (D) 2950 KHz
- 11. A 50% modulated AM is diode detected having time constant 5μ sec. The message frequency for which diagonal clippling will not occur is \_\_\_\_\_.
  (A) 1 MHz
  (B) 55 KHz
  - (C) 346 KHz (D) 28 KHz
- **12.** A PCM system uses a uniform quantizer followed by a 10bit binary encoder. The bit rate is 80Mbps. The

#### Time: 90 min.

maximum message bandwidth for which the system operates satisfactorily is \_\_\_\_\_.

- (A) 4 KHz (B) 4 MHz (C) 10 MHz (D) 10 KHz
- **13.** The function

$$F_{X,Y}(x,y) = b \left[ \pi + \tan^{-1} \left( \frac{x}{2} \right) \right] \left[ \pi + \tan^{-1} \left( \frac{y}{3} \right) \right]$$

is a valid joint distribution function for random variables *X* and *Y* if the constant b is \_\_\_\_\_.

(A) 
$$\frac{1}{\pi^2}$$
 (B)  $\frac{4}{\pi^2}$   
(C)  $\frac{4}{9\pi^2}$  (D)  $\frac{2}{3\pi^2}$ 

**14.** A joint sample space for two random variables *X* and *Y* has 3 elements (1, 2), (2, 3) and (3, 1). Probabilities of these elements are 0.2, 0.45 and 0.35 respectively.

The probability of the event 
$$\{X \le 2.5, Y \le 5\}$$
 is  
(A) 0.35 (B) 0.65 (C) 0.45

- (C) 0.45 (D) 0.2
- **15.** An amplitude modulated signal is represented by  $\phi_{AM}(t) = 10 \cos(2\pi \ 10^6 t) \ [1 + m(t)]$  where  $m(t) = 0.5 \cos(2\pi \ 10^3 t) + 0.2 \cos(4\pi 10^3 t)$  with regard to following which one of the following is true?
  - (1) Modulation index is 0.539
  - (2) Power of modulated signal is 57.25W
  - (3) Bandwidth of the modulated signal is 40 KHz
  - (4) Spectrum consist of 6 spectral components
  - (A) 1 and 2 (B) 2 and 3 (C) 2 and 4
  - (C) 2 and 4 (D) 1 and 4
- **16.** Two random variable *X* and *Y* have the probability density function

$$f_{XY}(x, y) = \begin{cases} \frac{xy}{12} & 0 < x < 3 \text{ and } 0 < y < 2\\ 0 & elsewhere \end{cases}$$

The X and Y are

- (A) Correlated but statistically dependent.
- (B) Uncorrelated but statistically independent.
- (C) Correlated but statistically independent.
- (D) Uncorrelated but statistically dependent.
- 17. Source  $S_1$  produces 4 discrete symbols with equal probability.

Source  $S_2$  produces discrete symbols with equal probability. If  $H_1$  and  $H_2$  are the entropies of sources  $S_1$  and  $S_2$  respectively, then which one of the following is correct?

- (A)  $H_1$  is always equal to  $H_2$
- (B)  $H_1$  is always less then  $H_2$
- (C)  $H_1$  is always greater than  $H_2$
- (D)  $H_2$  is 1.5 times  $H_1$  only

- 18. In a CD player, the sampling rate is 94.1 KHz and the samples are quantized using a 32 bit/sample quantizer. The resulting number of bits for a piece of music with a duration of 60 minutes are \_\_\_\_\_.
  (A) 108.4 × 10<sup>10</sup>bits (B) 10.84 × 10<sup>10</sup>bits
  - (C)  $1.084 \times 10^9$  bits (D)  $10.84 \times 10^9$  bits
- **19.** A discrete memoryless source have six symbols with probabilities

$$P_{A} = \frac{1}{4} \qquad P_{B} = \frac{1}{2}$$

$$P_{C} = \frac{1}{8} \qquad P_{D} = P_{E} = \frac{1}{20},$$

$$P_{F} = \frac{1}{40}$$

The amount of information contained in the message ABBAAB

(A)	6 bits	(B)	7 bits
(C)	8 bits	(D)	9 bits

**20.** A source delivers symbols  $X_1, X_2, X_3, X_4, X_5$  with probabilities  $\frac{1}{4} \frac{1}{4} \frac{1}{4} \frac{1}{8}$  and  $\frac{1}{8}$  respectively. The entropy of

	1110	0	
the s	system is		
(A)	3.32 bits/second	(B)	2.25 bits/message
(C)	2.25 bits/second	(D)	1.0 bits/message

**21.** A pseudo – noise sequence is generated using a feedback shift register of length m = 5. The chip rate is  $10^6$  chips/sec. The *PN* sequence length & chip duration is (A) 31 and 1 µs (B) 31 and 0.1 µs

(C) 
$$32 \text{ and } 1 \mu s$$
 (D)  $32 \text{ and } 0.1 \mu s$ 

- **22.** The period of *PN* sequence, if m = 6 and chip rate is  $10^7$  chips/sec \_\_\_\_\_.
  - (A) 6.4 μs
     (B) 6.3 μs
     (C) 63 μs
     (D) 64 μs
- 23. If in IS 95 CDMA system the base band information rate R = 4800 bps and total radio frequency bandwidth  $\omega = 2.5$ MHz & the number of users N = 18 then SNR at the base station receiver is \_\_\_\_\_.
  - (A) 30 dB (B) 14.86 dB
  - (C) 10 dB (D) 30.63 dB
- 24. An audio signal consisting of the sinusoidal term given<br/>as \_\_\_\_\_\_  $x(t) = 5\cos(500\pi t)$ The number of bits required to achieve signal -<br/>to quantization noise ratio of at least 30 dB.<br/>(A) 4 bits(A) 4 bits(B) 5 bits<br/>(C) 6 bits(D) 3 bits
- **25.** In a GSM system, which is a TDMA/FDD system 25 MHz frequency is used for the forward link, which can accommodate 200 KHz radio channels. If 8 speech channels are used & no guard band is assumed, then the total number of users in GSM are
  - (A) 125 users (B) 16 users
  - (C) 1000 users (D) 1600 users

- **26.** Consider the following statements about PCM
  - (1) PCM requires a large bandwidth as compared to other systems.
  - (2) Very high noise immunity.
  - (3) PCM is used in space communication.
  - (4) Repeaters regenerate the received PCM.
  - (A) 1, 2 & 3 are true (B) 2, 3, & 4 are true
- (C) Only 2 & 3 are true (D) Only 2 & 4 are true **27.** If a signal x(t) having PDF

$$f_x(x) = \begin{cases} 0 & , \text{ for } x < -4 \\ \frac{1}{8} & , \text{ for } -4 < x < 4 \\ 0 & , \text{ for } x > 4 \end{cases}$$

is quantized into 128 levels, then the nomalized signal power & noise power is

- (A) 5.33 W & 325.55 μW
  (B) 16 W & 3.25 mW
  (C) 5.33 W & 3.25 mW
  (D) 1 W & 325.55 μW
- **28.** A delta modulator system is designed to operate at 4 times the Nyquist rate for a signal with a 2 KHz bandwidth. The quantization step size is 36 mV. Maximum amplitude of a 0.5 KHz input sinusoid for which there is no slope overload is \_\_\_\_\_.

(A)	46 mV	(B)	183 mV
(C)	45.8 mV	(D)	184 mV

**29.** Matched filter output over (0, *T*) interval to the pulse wave form

$$x(t) = \begin{cases} e^{-2t} & ,0 \le t \le T \\ 0 & ,otherwise \end{cases}$$
(A)  $\frac{e^{-2T}}{2} \sin h(2t)$ 
(B)  $\frac{e^{-2T}}{2} \sin(2t)$ 
(C)  $e^{-2T} \sinh(2t)$ 
(D)  $e^{-2T} \sin(2t)$ 

**30.** If in a PCM system the number of quantization levels are decreased from 2048 to 8 then the bandwidth required for the transmission of PCM signal decreases by a factor of \_\_\_\_\_.

(A) 
$$\frac{11}{3}$$
 (B)  $\frac{1}{4}$   
(C)  $\frac{3}{11}$  (D)  $\frac{1}{3}$ 

**31.** If an RF pulse x(t) is applied at the input of the matched filter then the maximum magnitude of the signal at the

output of matched filter and the time at which it occurs are respectively.

$\operatorname{If} x(t) = \begin{cases} 10\sin(4\pi \times 10^6) \\ 0 \end{cases}$	)t volts	0.1  ms < t < 0.5  ms otherwise	
(A) 20 mV, 0.1 ms	(B)	80 mV, 0.5 ms	
(C) 20 mV, 0.5 ms	(D)	40 mV, 0.1 ms	

## Statement for Common data questions: 32 and 33:

A normal GSM has 3 start bits, 3 stop bits (also called as trailing bits), 26 training bits for allowing adaptive equalization, 8.25 guard bits and 2 bursts of 58 bits of encrypted data which is transmitted at 270.833 kbps in the channel.

**32.** Frame rate is

	(A) 1.733 frames/sec	(B)	1733.28 frames/sec
	(C) 216.66 frames/sec	(D)	21.666 frames/sec
33.	The frame efficiency is		
	(A) 54.24%	(B)	74.24%
	(C) 36.24%	(D)	59.36%

# Statement for linked questions 34 and 35:

Two messages are transmitted by space and mark using a simple binary pulse x(t)



Where A = 5V and  $T = 10^{-3}$ sec

**34.** The energy *E* of the signal x(t) is

(A) 4.22 W	(B)	8.33 mW
(C) 4.22 mW	(D)	8.33 W

**35.** If the probability of x(t) being present is 0.5 then the error probability of the optimum receiver is, if the

channel noise is white noise with PSD  $\frac{\eta}{2}$ , where  $n = 10^{-4}$ .

(A) 
$$\frac{1}{2} erfc \sqrt{\frac{250}{3}}$$
 (B)  $\frac{1}{2} erfc \sqrt{\frac{5}{6}}$   
(C)  $\frac{1}{2} erfc \sqrt{\frac{125}{3}}$  (D)  $\frac{1}{2} erfc \sqrt{\frac{500}{3}}$ 

Answer Keys									
1. A	<b>2.</b> D	<b>3.</b> D	<b>4.</b> B	<b>5.</b> A	<b>6.</b> A	<b>7.</b> D	<b>8.</b> B	<b>9.</b> B	<b>10.</b> C
11. B	12. B	13. C	14. B	15. A	16. A	17. D	18. D	19. D	<b>20.</b> B
21. A	<b>22.</b> B	<b>23.</b> B	<b>24.</b> B	<b>25.</b> C	<b>26.</b> B	27. A	<b>28.</b> B	<b>29.</b> A	<b>30.</b> C
31. C	<b>32.</b> C	<b>33.</b> B	<b>34.</b> B	35. A					

### HINTS AND EXPLANATIONS

1

1. As we know that in FSK  $f_i = \frac{n_c + i}{T_i}$ , for the signals to be orthogonal. Where  $T_{h}$  = bit duration  $n_{\rm o}$  = any fixed integer i = 1, 2 $\operatorname{So} f_1 = 25 \mathrm{KHz}$  $f_2 = 50 \text{KHz}$ by putting i = 1 & 2 we get  $T_{i} = 0.04$  ms Choice (A) 2.  $x(t) = 400 \cos(\omega c + 100\pi)t + 400 \cos(\omega - 100\pi)t +$  $2000 \cos \omega t$  $= 800 \cos(\omega_{c} t) \cos(100\pi t) + 2000 \cos \omega_{c} t$  $= [2000 + 800\cos(100\pi t)] \cos \omega_t t$ It is in the from of  $[A + f(t)] \cos \omega_t t$ Complex envelope is  $(2000 + 800 \cos(100\pi t))$ Choice (D) 3. Companding  $\rightarrow$  non – uniform quantization Squelch  $\rightarrow$  muting  $R_{\rm u}$ Double conversion  $\rightarrow$  Improving IRR Choice (D) 4. Choice (B) 5.  $m(t) = \operatorname{sinc}^2(4000t)$ So *m*(*t*) have *B*. *W* 4000Hz. and SSB – SC signal having B. W same as m(t) so — B. W. of SSB .SC = 4KHzChoice (A) 6. 1 1 0 1 1 0 0 is differential encoded So encoded signal is 0 1 1 1 0 0 1 0 (by taking reference 1) So transmited phase are  $\pi$ ,  $\pi$ ,  $\pi$ , 0, 0, 0,  $\pi$ , 0 So for 1 0 11 00 phase are  $(\pi, 0, 0, 0, \pi, 0)$ Choice (A) 7.  $\Rightarrow$  Bandwidth depends upon sampling rate  $\Rightarrow$  Cross – talk occurs in PAM. In PCM Inter symbol interference occurs. Bandwidth depends upon no. of quantization  $\Rightarrow$ levels  $\Rightarrow$ Quantization error = S/2, which depends upon step size Choice (D) 8. Choice (B) 9. As we know that Step size  $S = \frac{\text{Amplitude Range}}{\text{No. of quantization levels}} =$ and Quantization noise power  $S^2$  $A^2$ 

$$= \frac{1}{12} = \frac{1}{12 \times 16}$$
$$= \left(\frac{1}{192}\right)^{th} \text{ of total (amplitude range)}^2$$

Choice (B)

10. Double spotting occurs at image signal so Receiver should be tuned at =  $2500 - 2 \times 450 = 1600$  KHz Choice (C)

1. 
$$m_a = 0.5$$
  
 $\frac{1}{RC} \ge \frac{\omega_m m_a}{\sqrt{1 - m_a^2}}$  where  $RC = 5\mu$  sec  
 $\frac{0.2 \times 10^6 \times .86}{0.5} \ge \omega_m$ 

$$\omega_m = 0.346 \times 10^6$$
$$= 55 \text{KHz}$$

12. Bit rate = 80 Mbps So Nyquist Rate =  $\frac{80Mbps}{10bit}$  = 8MHz

So message bandwidth = 
$$\frac{8}{2}$$
 MHz = 4MHz

Choice (B)

**13.** 
$$F_{\chi\gamma}(\infty, \infty) = b \left[ \pi + \frac{\pi}{2} \right] \left[ \pi + \frac{\pi}{2} \right]$$
 should be 1  

$$\Rightarrow b \left[ \frac{3\pi}{2} \right] \left[ \frac{3\pi}{2} \right] = 1$$

$$b \frac{9\pi^2}{4} = 1; \ b = \frac{4}{9\pi^2}$$
Choice (C)

- **14.**  $F_{XY}(x, y) = 0.2u(x 1) u(y 2) + 0.45 u(x 2) u(y 3) + 0.35u(x 3) u(y 1)$ and probability that  $X \le 2.5$ ,  $Y \le 5$  is = 0.2 + 0.45 = 0.65 Choice (B)
- **15.**  $\phi_{AM}(t) = 10 \cos(2\pi \times 10^6)t [1 + m(t)]$   $m(t) = 0.5 \cos(2\pi 10^3 t) + 0.2\cos(4\pi \times 10^3 t)$   $\phi_{AM}(t) = 10 \cos(2\pi \times 10^6)t [1 + 0.5 \cos(2\pi 10^3 t) + 0.2 \cos(4\pi \times 10^3 t)]$ It is a multi – tone modulation so  $m_{a1} = 0.5$  $m_{a2} = 0.2$

$$m_a^{a2} = \sqrt{m_{a1}^2 + m_{a2}^2} = \sqrt{0.25 + 0.04} = 0.539$$

Power of modulated signal is

$$P = \frac{10^2}{2} \left( 1 + \frac{m_1^2 + m_2^2}{2} \right)$$
$$= 50 \left( 1 + \frac{0.25 + 0.04}{2} \right) = 57.25 \text{ W}$$

Spectrum of modulated signal is



Where 
$$f_c = 1$$
 MHz  
 $f_1 = 1$  KHz  
 $f_2 = 2$  KHz  
So bandwidth =  $2f_2 = 4$  KHz  
and 5 spectral components are present. Choice (A)

16. 
$$R_{XY} = \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} xy \ f_{XY}(x, y) dx \ dy$$
  
 $= \int_{0}^{3} \int_{0}^{2} \frac{x^{2} y^{2}}{12} dx \ dy = 2$   
 $E[X] = \int_{0}^{3} \int_{0}^{2} \frac{x^{2} y}{12} dx \ dy$   
 $= \frac{1}{12} \left[ \frac{x^{3}}{3} \right]_{0}^{3} \left[ \frac{y^{2}}{2} \right]_{0}^{2} = \frac{3}{2}$   
 $E[Y] = \int_{0}^{3} \int_{0}^{2} \frac{x \ y^{2}}{12} dx \ dy$   
 $= \left[ \frac{x^{2}}{2} \right]_{0}^{3} \left[ \frac{y^{3}}{3} \right]_{0}^{2} \frac{1}{12} = 1$ 

Since  $R_{XY} = 2$ , &  $E[X] E[Y] = \frac{3}{2} \times 1 = \frac{3}{2}$ 

 $R_{XY} \neq E[X] E[Y]$ 

So X and Y are in correlated form and Marginal densities are

$$f_x(x) = \int_0^2 \frac{xy}{12} dy = \frac{x}{6}, \ 0 < x < 3$$
  
$$f_y(y) = \int_0^3 \frac{xy}{12} dx = \frac{3y}{8}, \ 0 < y < 2$$
  
so  $f_x(x) f_y(y) = \frac{3xy}{48}, \ 0 < x < 3$  and  $0 < y < 2$   
$$= \frac{xy}{16}$$

 $f_{XY}(x, y) \neq f_x(x) f_y(y)$ so X and Y are statistically dependent.

**17.** 
$$H(x) = \sum_{i=1}^{n} P_i \cdot \log_2 \frac{1}{P_i}$$
  
 $H_1 = 4 \times \frac{1}{4} \cdot \log_2^4 = 2$   
 $H_2 = 8 \times \frac{1}{8} \cdot \log_2^8 = 3$   
 $H_2 = 1.5 H_1$  Choice (D)  
**18.** Sampling rate is  $f = 94.1 \times 10^3$ 

= 94100 samples/sec

each sample is quantized using 32 - bitsTotal number of bits/ sec =  $94100 \times 32 = 3011200bits$ For a music piece of duration 60min The number of bits/channel is =  $94100 \times 32 \times 60 \times 60 = 10.84 \times 10^9$  bits Choice (D)

**19.** 
$$P(ABBAAB) = \frac{1}{4} \left(\frac{1}{2}\right)^2 \times \left(\frac{1}{4}\right)^2 \times \frac{1}{2} = \frac{1}{2^9}$$
  
 $I = \log_2 \frac{1}{P} = \log_2^{2^9} = 9$  bits. Choice (D)

**20.** Entropy 
$$H(X) = -[0.25\log_2 0.25 + 0.25\log_2 0.25 + 0.25\log_2 0.25 + 0.125\log_2 0.125 + 0.125\log_2 0.125]$$
  
= 2.25 bits/message Choice (B)

**21.** The PN sequence length is  $N = 2^m - 1 = 2^5 - 1 = 31$ The chip duration is  $T_c = \frac{1}{10^6} = 1 \mu s$ . Choice (A)

**22.** The period of the PN sequence is 
$$1$$

$$T = NT_c = (2^6 - 1) \times \frac{1}{10^7} = 6.3 \ \mu s$$
 Choice (B)

23. 
$$\frac{E_b}{N_o} = \frac{W/R}{N-1}$$
  
=  $\frac{2.5 \times 10^6 / 4800}{(18-1)} = 14.86 \text{ dB}$  Choice (B)

24. We know that  $\left(\frac{S}{N}\right)_{dB} = 1.8 + 6N$ Where N = No. of bits used to encode one sample in PCMTo get SNR at least 30 dB  $1.8 + 6N \ge 30$  dB  $6N \ge 28.2$  $N \ge 4.7 \approx 5$  bits Choice (B)

25. 
$$N = \frac{25 \text{MHz}}{(200 \text{KHz})/8} = 1000$$
 Choice (C)

**26.** Choice (B)

Choice (A)

27. mean square value of a random variable X is given as  $\overline{X^2} = \int_{-\infty}^{+\infty} x^2 f_x(x) dx$ 

So mean square value of x will be

$$\overline{x^{2}} = \int_{-4}^{+4} x^{2} \cdot \frac{1}{8} dx = \frac{1}{8} \left[ \frac{X^{3}}{3} \right]_{-4}^{+4} = \frac{16}{3}$$
  
Signal power =  $\frac{\overline{x^{2}}}{R}$ 

Normalized signal power =  $\frac{x^2}{1} = \frac{16}{3}$  W

and We know that

Step size 
$$S = \frac{2x_{max}}{No.of \ quantization \ levels}$$
  

$$= \frac{2 \times 4}{128} = \frac{1}{16} V$$
Nomalized signal power  $= \frac{S^2}{12} = \frac{1}{16} \times \frac{1}{16 \times 12}$ 

$$= 325.55 \ \mu W$$
Choice (A)  
**28.** Sinusoid input m(t) =  $A \cos \omega_m t$   
Slope of the input  $= \frac{dm(t)}{dt}$   
To avoid slope overload  $\frac{S}{T_s} \ge \left|\frac{dm(t)}{dt}\right|_{max}$ ;  $\frac{S}{T_s} \ge A\omega_m$   
 $A_{max} = \frac{S}{T_s \omega_m}$   
Where  $S = 36 \times 10^{-3} V$   
 $\omega_m = 2\pi \times 0.5 \times 10^3$   
 $T_s = \frac{1}{f_s} = \frac{1}{2 \times 4 \times 2 \times 10^3}$   
 $A_{max} = \frac{36 \times 10^{-3}}{2\pi \times 0.5 \times 10^3} .4 \times 2 \times 2 \times 10^3$   
 $= 183 \ mV$ 
Choice (B)

**29.** The impulse response of the matched filter is  $h(t) = x(T-t) = e^{-2(T-t)} = e^{-2T+2t}$ Now output  $y(t) = x(t) \otimes h(t)$ 

$$\Rightarrow y(t) = \int_{-\infty}^{+\infty} e^{-2\tau} \cdot e^{-2(T-t+\tau)} d\tau$$
$$= e^{-2T} \int_{-\infty}^{+\infty} e^{-2\tau} \cdot e^{(2t-2\tau)} d\tau = e^{-2t} \int_{-\infty}^{+\infty} e^{(2t-4\tau)} d\tau$$
$$= \frac{e^{-2T}}{-4} \left[ e^{2t-4\tau} \right]_{0}^{t} = \frac{e^{-2T}}{-4} \left[ e^{-2t} - e^{2t} \right]$$
$$= \frac{e^{-2T}}{2} \left[ \frac{e^{2t} - e^{-2t}}{2} \right] = \frac{e^{-2T}}{2} \sinh(2t) \quad \text{Choice (A)}$$

- **30.** For 2048 quantization levels, the number of bits required =  $2048 = 2^N$ 
  - So N = 11
  - for 8 quantization levels

Number of bits required to encode one sample =  $8 = 2^3 = N = 3$ 

Bandwidth depends upon number of bits required to encode one sample.

So decresed ratio = 
$$\frac{3}{11}$$
 Choice (C)

**31.** As we know that in matched filter the maximum amplitude occurs at t = T and It is equal to E So T = 0.5ms

Energy of 
$$x(t) = \int_{-\infty}^{+\infty} x^2(t) dt = \int_{0.1}^{0.5} 10^2 \sin^2(4\pi \times 10^6) t dt$$
  
=  $\frac{10 \times 10}{2} \int_{0.1}^{0.5} \left[ 1 - \cos(8\pi \times 10^6) t \right] dt$   
=  $\frac{10 \times 10}{2} [0.5 - 0.1] - 0 = 20 \text{ mV}$  Choice (C)

- 32. A time slot has 6 + 8.25 + 26 + 2(58) = 156.25 bits Since in GSM each frame consists of 8 time slots So number of bits/frame =  $156.25 \times 8 = 1250$  bits/frame So frame rate =  $\frac{270.833$ kpbs}{1250bits/frame} = 216.66 frame/sec Choice (C)
- 33. The total number of over head bits in on frame = 8(6) + 8(8.28) + 8(26) = 322 bits Frame efficiency =  $\eta = 1 - \frac{322}{1250} = 74.24\%$

34. 
$$x(t) = \begin{cases} \frac{2A}{T}t , 0 \le t \le T/2 \\ 2A - \frac{2A}{T}t , T/2 \le t \le T \end{cases}$$
  
Energy  $E = \int_{-\infty}^{+\infty} x^{2}(t)dt$   
 $E = \int_{0}^{\frac{T}{2}} \left(\frac{2A}{T}t\right)^{2} dt + \int_{\frac{T}{2}}^{T} \left(2A - \frac{2A}{T}t\right)^{2} dt$   
 $= \frac{4A^{2}}{T^{2}} \int_{0}^{\frac{T}{2}} t^{2} dt + 4A^{2} \int_{\frac{T}{2}}^{T} \left(1 - \frac{t}{T}\right)^{2} dt$   
 $= \frac{4A^{2}}{T^{2}} \left[\frac{t^{3}}{3}\right]_{0}^{\frac{T}{2}} + 4A^{2} \left[t - \frac{2t^{2}}{T^{2}} + \frac{1}{T^{2}} \frac{t^{3}}{3}\right]_{\frac{T}{2}}^{T}$   
 $= \frac{A^{2}T}{6} + \frac{4A^{2}T}{24} = \frac{A^{2}T}{3}$   
 $E = \frac{5 \times 5 \times 10^{-3}}{3} = 8.33 \text{ mW}$  Choice (B)

**35.** Since channel noise is white noise. So the optimum filter is same as matched filter. So probability of error of matched filter is

$$P^{e} = \frac{1}{2} \operatorname{erfc} \sqrt{\frac{E}{\eta}} = \frac{1}{2} \operatorname{erfc} \sqrt{\frac{25 \times 10^{-3}}{3 \times 10^{-4}}}$$
$$= \frac{1}{2} \operatorname{erfc} \sqrt{\frac{250}{3}}$$
Choice (A)