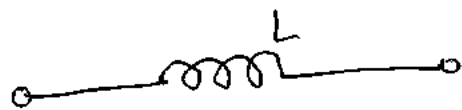


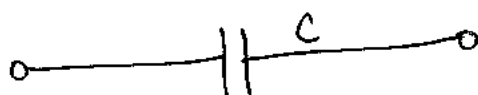
Ch-3 State Space Analysis

- State gives future behaviour of the system based on present input and past history of system.
Past history = (initial condition) of system is described by state variables.
- State of dynamic system is nothing but minimum set of variables such that knowledge of this state variable at $t = t_0$ and with knowledge of their inputs at for $t > t_0$ we can get future behaviour of system for $t > t_0$.
- Resistive ckt don't have any state variable because present output doesn't depend on past history of system. So in resistive ckt present output depends only on present input. Resistive ckt can't store any energy.

So i.e. no past history $\Rightarrow$ NO STATE VARIABLES. So pure resistive ckt is called STATIC/MEMORYLESS



$I_L \rightarrow$ state variable



$V_C \rightarrow$ state variable

Transfer function approach

- ① LTI system only
- ② Initial conditions are zero
- ③ Internal states of the system are not taking into account

State space Analysis

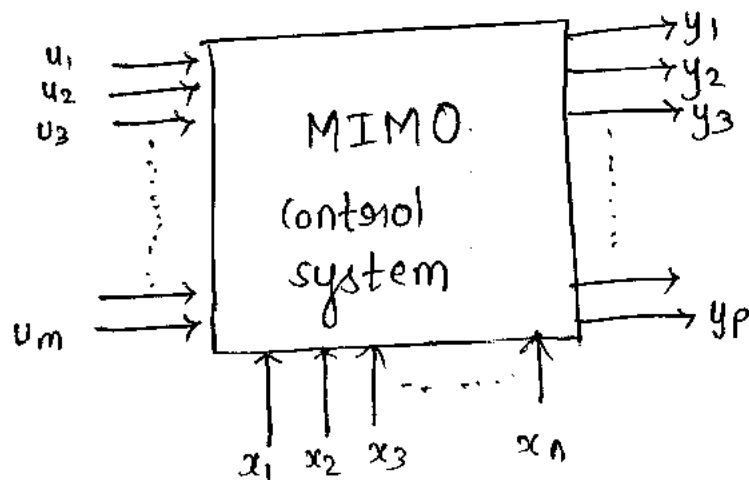
- ① Non-linear & linear time invariant and time variant system
- ② Initial conditions are not zero
- ③ Internal states of the system are taking into account

Advantages of state space analysis

- State space analysis is applicable to both linear, non-linear and time invariant, ^{Time variant} system
- whereas T.F. approach is applicable to LTI system only
- I.C. are taking into account whereas T.F. approach doesn't.
- The internal states of the system can also be determine by this approach unlike transfer function approach.
- Analysis of multiple input & multiple output is easy with this approach but by using T.F. approach it is very cumbersome.

-Controllability and observability can also be determined by this approach ~~an~~ unless transfer function approach.

*State space representation



m : number of inputs

p : number of output

n : number of state variable

*Standard form

$$\dot{X} = A \cdot X + B \cdot U \quad \text{--- (1)}$$

$\nearrow$ differential state variable

$$Y = C \cdot X + D \cdot U \quad \text{--- (2)}$$

$\downarrow$ output vector $\downarrow$ state variable $\downarrow$ input vector

A = state matrix

B = input matrix

C = output matrix

D = transmittance matrix

$$U = \begin{bmatrix} u_1 \\ u_2 \\ \vdots \\ u_m \end{bmatrix}_{m \times 1}$$

$$Y = \begin{bmatrix} y_1 \\ y_2 \\ y_3 \\ \vdots \\ y_p \end{bmatrix}_{p \times 1}$$

$$\dot{X}_{n \times 1} = A_{n \times n} X_{n \times 1} + B_{n \times m} U_{m \times 1}$$

$$Y_{n \times 1} = C_{p \times n} X_{n \times 1} + D_{p \times m} U_{m \times 1}$$

$$\dot{X} = \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \vdots \\ \dot{x}_n \end{bmatrix}_{n \times 1}$$

- If RLC circuit is given then no. of state variables is equal to no. of memory storage elements

- If differential eqⁿ is given the no. of state variables is equal to no. of order of diff. eqⁿ

* State space representation to differential eqⁿ

Q:- Find state model to following system?

$$\ddot{y} + 3\dot{y} + 5y = 10U$$

No. of state variables = Order of diff. eqⁿ

No. of state variable = 3

$\downarrow$
 x_1, x_2, x_3

→ Differential state variable $\Rightarrow \dot{x}_1, \dot{x}_2, \dot{x}_3$

$$\text{Let } \boxed{y = x_1}$$

$$\dot{y} = \frac{dx_1}{dt} = \boxed{\dot{x}_1 = x_2}$$

$$\ddot{y} = \boxed{\dot{x}_2 = x_3}$$

$$\ddot{\ddot{y}} = \dot{x}_3$$

$$\dot{X} = AX + BU$$

$$Y = CX + DU$$

$$\dot{x}_3 + 3x_3 + 5x_2 + 7x_1 = 10U$$

$$\boxed{\dot{x}_3 = -7x_1 - 5x_2 - 3x_3 + 10U}$$

$$\dot{x}_1 = 0x_1 + 0x_2 + 0x_3 + 0U$$

$$\dot{x}_2 = 0x_1 + 0x_2 + x_3 + 0U$$

$$\dot{x}_3 = -7x_1 - 5x_2 - 3x_3 + 10U$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix}_{3 \times 1} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -7 & -5 & -3 \end{bmatrix}_{3 \times 3} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}_{3 \times 1} + \begin{bmatrix} 0 \\ 0 \\ 10 \end{bmatrix}_{3 \times 1} U$$

$$y_{1 \times 1} = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix}_{1 \times 3} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}_{3 \times 1}$$

Short-cut

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 & \dots & 0 \\ 0 & 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 0 & 1 & \dots & 0 \\ \vdots & & & & \ddots & \\ 0 & 0 & 0 & 0 & \dots & 1 \end{bmatrix}_{n \times n}$$

co-efficient from $L \rightarrow R^F$ with
opposite sign last first

$$B = \begin{bmatrix} 0 \\ 0 \\ 0 \\ \vdots \\ k \end{bmatrix}$$

$k \rightarrow$ co-efficient of u

$$C = [1 \ 0 \ 0 \ \dots \ 0]$$

Q:- $\ddot{y} + 2\ddot{y} + 4\ddot{y} + 6\ddot{y} + 8\ddot{y} + 10y = 5u$

No. of state variable = 5

↓

$$x_1, x_2, x_3, x_4, x_5$$

let $y = x_1$

$$\dot{y} = \dot{x}_1 = x_2$$

$$\ddot{y} = \dot{x}_2 = x_3$$

$$\ddot{\ddot{y}} = \dot{x}_3 = x_4$$

$$\ddot{\ddot{\ddot{y}}} = \dot{x}_4 = x_5$$

$$\ddot{\ddot{\ddot{\ddot{y}}}} = \dot{x}_5$$

$$\dot{x}_5 = -2x_5 - 4x_4 - 6x_3 - 8x_2 - 10x_1 + 5u$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \\ \dot{x}_4 \\ \dot{x}_5 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 1 \\ -10 & -8 & -6 & -4 & -2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \\ x_5 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 5 \end{bmatrix} U$$

$$y = [1 \ 0 \ 0 \ 0 \ 0]_{1 \times 5} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \\ x_5 \end{bmatrix}_{5 \times 1} + 0$$

State models are not unique

There are 4 different types of state model

1. Controllable canonical form CCF

2. Observable canonical form OCF

3. Diagonalization OR Normal form

4. Jordan canonical form

Above form is controllable canonical form.

Q:- How to find OCF from CCF

$$A_{OCF} = [A_{CCF}]^T$$

$$B_{OCF} = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 5 \end{bmatrix} \begin{matrix} \text{End} \\ \uparrow \\ \text{Start} \end{matrix} = \begin{bmatrix} 5 \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix}_{OCF}$$

$$C_{ccf} = [1 \ 0 \ 0 \ 0 \ 0]_{1 \times 5}$$

end ← start

$$C_{ocf} = [0 \ 0 \ 0 \ 0 \ 1]_{1 \times 5}$$

* State model to the transfer function

Q:- Write state model to the given transfer function?

$$\frac{Y[s]}{U[s]} = \frac{2s+3}{s^2+5s+6}$$

$$\left(\frac{Y[s]}{X[s]} \right) \left(\frac{X[s]}{U[s]} \right) = \frac{2s+3}{s^2+5s+6}$$

$$\frac{Y[s]}{X[s]} = 2s+3 = N(s)$$

$$; \frac{X[s]}{U[s]} = \frac{1}{s^2+5s+6} = \frac{1}{D(s)}$$

$$Y[s] = 2sX[s] + 3X[s]$$

$$y = 2\dot{x} + 3x$$

$$\ddot{x} = \dot{x}_2$$

$$\dot{x} = \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} x_2 \\ \ddot{x} \end{bmatrix}$$

$$x = \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

$$y = 2x_2 + 3x_1$$

$$\dot{x}_2 = -6x_1 - 5x_2 + u$$

$$s^2 X[s] + 5s X[s] + 6X[s] = U[s]$$

$$\ddot{x} + 5\dot{x} + 6x = u$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix}_{2 \times 1} = \begin{bmatrix} 0 & 1 \\ -6 & -5 \end{bmatrix}_{2 \times 2} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$$

$$y = \begin{bmatrix} 3 & 2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

Short-cut

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 & \dots & 0 \\ 0 & 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 0 & 1 & \dots & 0 \\ \vdots & & & & \ddots & \\ 0 & 0 & 0 & 0 & \dots & 1 \end{bmatrix}_{n \times n}$$

[co-efficient of $D(s)$ from Last $\rightarrow$ First with opp. sign]

$$C = [\text{co-efficient of } N(s) \text{ from } \frac{L}{D} \rightarrow F \text{ with same sign}]_{1 \times p}$$

$$B = \begin{bmatrix} 0 \\ 0 \\ 0 \\ \vdots \\ k \end{bmatrix}$$

NOTE:-

$$s^n = \dot{x}_n$$

$$Q:- \frac{Y[s]}{U[s]} = \frac{2s^3 + 4s^2 + 6}{s^5 + 3s^4 + 5s^3 + 7s^2 + 9s + 10}$$

$$\frac{Y[s]}{X[s]} \cdot \frac{X[s]}{U[s]} = \frac{2s^3 + 4s^2 + 6}{s^5 + 3s^4 + 5s^3 + 7s^2 + 9s + 10}$$

$$\frac{Y[s]}{X[s]} = \frac{2s^3 + 4s^2 + 6}{s^5 + 3s^4 + 5s^3 + 7s^2 + 9s + 10}$$

$$Y[s] = 2s^3 X[s] + 4s^2 X[s] + 6X[s]$$

$$y = 2\ddot{x} + 4\ddot{x} + 6x$$

$$s^5 + 3s^4 + 5s^3 + 7s^2 + 9s + 10$$

$$\cancel{L[X]} \quad \cancel{Y[X]}$$

$$\ddot{x} = \dot{x}_2 = x_3, \quad \ddot{x} = \dot{x}_3 = x_4$$

$$\dot{x} = \dot{x}_1 = x_2$$

$$x = x_1$$

$$\therefore y = 2x_4 + 4x_3 + 6x_1$$

$$\frac{X[s]}{U[s]} = \frac{1}{s^5 + 3s^4 + 5s^3 + 7s^2 + 9s + 10}$$

$$X[s]s^5 + 3s^4 X[s] + 5s^3 X[s] + 7s^2 X[s] + 9s X[s] + 10X[s] = U[s]$$

$$\ddot{\ddot{x}} + 3\ddot{\ddot{x}} + 5\ddot{x} + 7\ddot{x} + 9\dot{x} + 10x = 0$$

$$\dot{x}_5 + 3\dot{x}_5 + 5x_4 + 7x_3 + 9x_2 + 10x_1 = 0$$

$$\dot{x}_5 = 0 - 3x_5 - 5x_4 - 7x_3 - 9x_2 - 10x_1$$

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 1 \\ -10 & -3 & -7 & -5 & -3 \end{bmatrix}_{5 \times 5}$$

$$B = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 2 \end{bmatrix}_{5 \times 1}$$

$$C = [3 \ 0 \ 2 \ 1 \ 0]_{1 \times 5}$$

$$Q:- \frac{Y[s]}{U[s]} = \frac{1}{(s+1)(s+2)(s+3)}$$

$$\frac{Y[s]}{U[s]} = \frac{k_1}{s+1} + \frac{k_2}{s+2} + \frac{k_3}{s+3}$$

$$\frac{Y[s]}{U[s]} = \frac{1/2}{s+1} + \frac{-1}{s+2} + \frac{1/2}{s+3}$$

$$Y[s] = \frac{U[s]}{2(s+1)} + \frac{(-1)U[s]}{s+2} + \frac{U[s]}{2(s+3)}$$

$$Y[s] = \underbrace{\frac{1/2 U[s]}{s+1}}_{x_1} + \underbrace{\frac{(-1)U[s]}{s+2}}_{x_2} + \underbrace{\frac{1/2 U[s]}{s+3}}_{x_3}$$

$$\therefore y = x_1 + x_2 + x_3$$

$$X_1 = \frac{\frac{1}{2} u[s]}{s+1}$$

$$sX_1 + X_1 = \frac{1}{2} u[s]$$

$$\dot{X}_1 + X_1 = \frac{1}{2} u$$

$$\dot{X}_1 = -X_1 + \frac{1}{2} u$$

$$X_2 = \frac{(-1) u[s]}{s+2}$$

$$sX_2 + 2X_2 = -u(s)$$

$$\dot{X}_2 + 2X_2 = -u$$

$$\dot{X}_2 = -u - 2X_2$$

$$X_3 = \frac{\frac{1}{2} u[s]}{s+3}$$

$$\dot{X}_3 = -3X_3 + \frac{1}{2} u$$

$$\begin{bmatrix} \dot{X}_1 \\ \dot{X}_2 \\ \dot{X}_3 \end{bmatrix} = \begin{bmatrix} -1 & 0 & 0 \\ 0 & -2 & 0 \\ 0 & 0 & -3 \end{bmatrix} \begin{bmatrix} X_1 \\ X_2 \\ X_3 \end{bmatrix} + \begin{bmatrix} \frac{1}{2} \\ -1 \\ \frac{1}{2} \end{bmatrix} u$$

$$Y = \begin{bmatrix} 1 & 1 & 1 \end{bmatrix} \begin{bmatrix} X_1 \\ X_2 \\ X_3 \end{bmatrix}$$

C

no. of poles
location

Partial fraction
co-efficient

Matrix B & C can be interchanged in diagonalization form.

$$Q:- \frac{Y[s]}{U[s]} = \frac{2(s+1)}{(s+3)(s+4)(s+6)}$$

$$= \frac{k_1}{s+3} + \frac{k_2}{s+4} + \frac{k_3}{s+6}$$

$$= \frac{-4/3}{s+3} + \frac{3}{s+4} + \frac{-5/3}{s+6}$$

$$\frac{2(-2)}{(1)(3)}$$

$$\frac{-3 \cdot 2}{-1(2)}$$

$$\frac{2(-5)}{(-2)(-3)}$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} -3 & 0 & 0 \\ 0 & -4 & 0 \\ 0 & 0 & -6 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} -4/3 \\ 3 \\ -5/3 \end{bmatrix} u$$

A B

$$Y = \begin{bmatrix} 1 & 1 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

C

* Jordan canonical form

$$Q:- \frac{Y[s]}{U[s]} = \frac{1}{(s+2)^2(s+3)}$$

$$\frac{Y[s]}{U[s]} = \frac{k'}{(s+2)^2} + \frac{k''}{s+3}$$

$$\frac{1}{(s+2)^2(s+3)} = \frac{k_1}{s+2} + \frac{k_2}{(s+2)^2} + \frac{k_3}{s+3}$$

$$1 = k_1(s+2)(s+3) + k_2(s+3) + k_3(s+2)^2$$

$$1 = k_2(1)$$

$$\boxed{k_2 = 1}$$

$$\boxed{1 = k_3}$$

$$1 = 6k_1 + 3k_2 + 4k_3$$

$$\boxed{k_1 = -1}$$

$$\frac{Y[s]}{U[s]} = \frac{-1}{s+2} + \frac{1}{(s+2)^2} + \frac{1}{(s+3)}$$

$$\frac{Y[s]}{U[s]} = 1 \cdot \underbrace{\frac{U[s]}{(s+2)^2}}_{x_1} - 1 \cdot \underbrace{\frac{U[s]}{s+1}}_{x_2} + 1 \cdot \underbrace{\frac{U[s]}{s+3}}_{x_3}$$

~~$$Y = k_1 x_1 + k_2 x_2 + k_3 x_3$$~~

$$Y = k_1 x_1 + k_2 x_2 + k_3 x_3$$

$$x_2 = \frac{U[s]}{s+2}$$

$$\dot{x}_2 = -2x_2 + U$$

$$x_1 = \frac{U[s]}{(s+2)^2} = \underbrace{\frac{U[s]}{(s+2)}}_{x_2} \cdot \frac{1}{(s+2)}$$

$$x_1 = x_2 \cdot \frac{1}{(s+2)}$$

$$s x_1 + 2 x_1 = x_2$$

$$\dot{x}_1 = -2x_1 + x_2$$

$$\dot{x}_3 = -3x_3 + U$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \underbrace{\begin{bmatrix} -2 & 1 & 0 \\ 0 & -2 & 0 \\ 0 & 0 & -3 \end{bmatrix}}_A \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \underbrace{\begin{bmatrix} 0 \\ 1 \\ 1 \end{bmatrix}}_B U$$

$$Y = \underbrace{\begin{bmatrix} 1 & -1 & 1 \end{bmatrix}}_C \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

Partial fraction coefficient

$$Q: \frac{y[s]}{u[s]} = \frac{1}{(s+5)^3(s+3)}$$

$$A = \begin{bmatrix} -5 & 1 & 0 & 0 \\ 0 & -5 & 1 & 0 \\ 0 & 0 & -5 & 0 \\ 0 & 0 & 0 & -3 \end{bmatrix}_{4 \times 4}, \quad B = \begin{bmatrix} 0 \\ 0 \\ 1 \\ 1 \end{bmatrix}$$

$$C = [k_1 \ k_2 \ k_3 \ k_4] = \left[-\frac{1}{2} \quad -\frac{1}{4} \quad -\frac{1}{8} \quad \frac{1}{8}\right]$$

$$\frac{1}{(s+5)^3(s+3)} = \frac{k_1}{(s+5)^3} + \frac{k_2}{(s+5)^2} + \frac{k_3}{(s+5)} + \frac{k_4}{s+3}$$

$$1 = k_1(s+3) + k_2(s+5)(s+3) + k_3(s+5)^2(s+3) + k_4(s+5)^3$$

$$\text{At } s = -3$$

$$\text{At } s = -5$$

Co-efficient of s^3

$$1 = k_4(2)^3$$

$$1 = k_1(-2)$$

$$0 = k_3 + k_4$$

$$\boxed{k_4 = \frac{1}{8}}$$

$$\boxed{k_1 = -\frac{1}{2}}$$

$$\boxed{k_3 = -\frac{1}{8}}$$

$$\text{At } s = 0$$

$$1 = 3k_1 + 15k_2 + 75k_3 + 125k_4$$

$$1 = 3\left(-\frac{1}{2}\right) + 15k_2 + 75\left(-\frac{1}{8}\right) + 125 \cdot \frac{1}{8}$$

$$1 = -\frac{3}{2} + 15k_2 + \frac{50}{8}$$

$$1 = \frac{-6 + 25}{4} + 15k_2$$

$$1 = \frac{19}{4} + 15k_2$$

$$\boxed{k_2 = -\frac{1}{4}}$$

* State Model to SFG (signal flow graph)

$$\frac{1}{s} \rightarrow \text{term} \rightarrow \boxed{\int}$$

$$\boxed{\int} \rightarrow 1/s$$

$$\boxed{d/dt} \rightarrow s$$

$$\Rightarrow \dot{x}_1 \xrightarrow{1/s} x_1$$

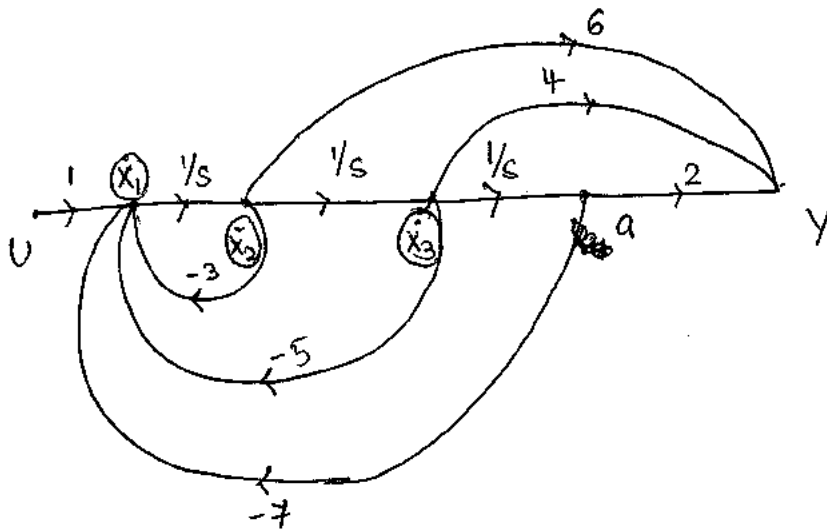
$$\dot{x} = \frac{d}{dt} x$$

$$\int \frac{d}{dt} x dt = x$$

NOTE:-

$\Rightarrow$ To select a node as a state variable the incoming branch ~~has~~ to particular node must be integrator

Q:-



For state variable, no. of $1/s$ terms = no. of state variable

$$\dot{x}_1 = -3\dot{x}_2 - 5\dot{x}_3 - 7U + U$$

$$\dot{x}_2 = \frac{1}{s} \dot{x}_1 = x_1$$

$$\boxed{\dot{x}_2 = x_1}$$

$$\dot{X}_3 = \frac{1}{S} \dot{X}_2 = X_2$$

$$\dot{X}_3 = X_2$$

$$a = \dot{X}_3 \cdot \frac{1}{S} = X_3$$

$$\dot{X}_1 = -3X_1 - 5X_2 - 7X_3 + U$$

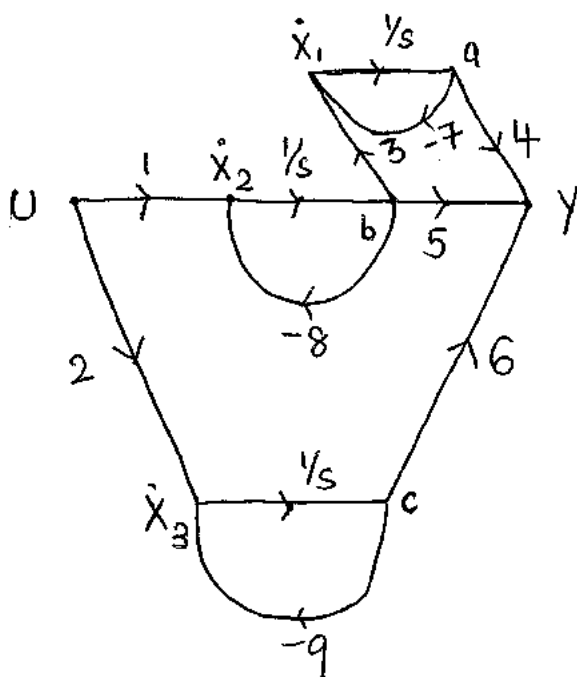
$$Y = 2a + 4\dot{X}_3 + 6\dot{X}_2$$

$$Y = 6X_1 + 4X_2 + 2X_3$$

$$\begin{bmatrix} \dot{X}_1 \\ \dot{X}_2 \\ \dot{X}_3 \end{bmatrix} = \underbrace{\begin{bmatrix} -3 & -5 & -7 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix}}_A \begin{bmatrix} X_1 \\ X_2 \\ X_3 \end{bmatrix} + \underbrace{\begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}}_B U$$

$$Y = \underbrace{[6 \ 4 \ 2]}_C \begin{bmatrix} X_1 \\ X_2 \\ X_3 \end{bmatrix}$$

Q:-



$$\dot{X}_1 = -7X_1 + 3b$$

$$\dot{X}_1 = -7X_1 + 3X_2$$

$$a = X_1$$

$$b = X_2$$

$$c = X_3$$

$$\dot{x}_2 = u - 8x_2$$

$$y = 5b + 4a + 6c$$

$$y = 5x_2 + 4x_1 + 6x_3$$

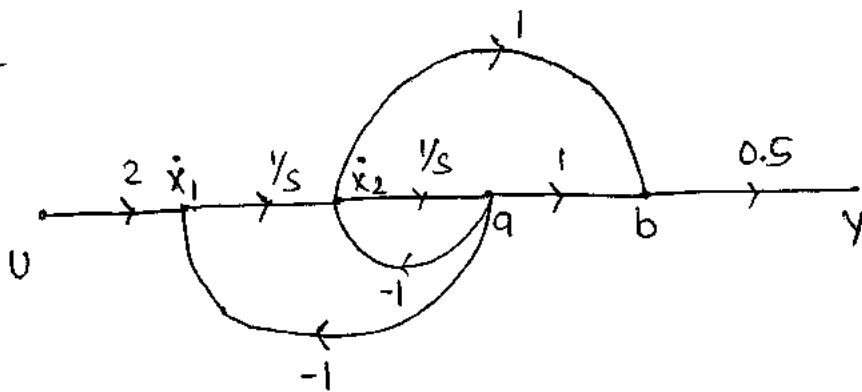
$$\dot{x}_3 = 2u - 9c$$

$$\dot{x}_3 = 2u - 9x_3$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \underset{A}{\begin{bmatrix} -7 & 8 & 0 \\ 0 & -8 & 0 \\ 0 & 0 & -9 \end{bmatrix}} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \underset{B}{\begin{bmatrix} 0 \\ 1 \\ 2 \end{bmatrix}} u$$

$$y = \underset{C}{\begin{bmatrix} 4 & 5 & 6 \end{bmatrix}} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

Q:-



$$\dot{x}_1 = 2u - a$$

$$\dot{x}_1 = 2u - x_2$$

$$\dot{x}_2 = x_1 - a$$

$$\dot{x}_2 = x_1 - x_2$$

$$y = 0.5b$$

$$y = 0.5x_1$$

$$a = \dot{x}_2 \cdot \frac{1}{5} = x_2$$

$$a = x_2$$

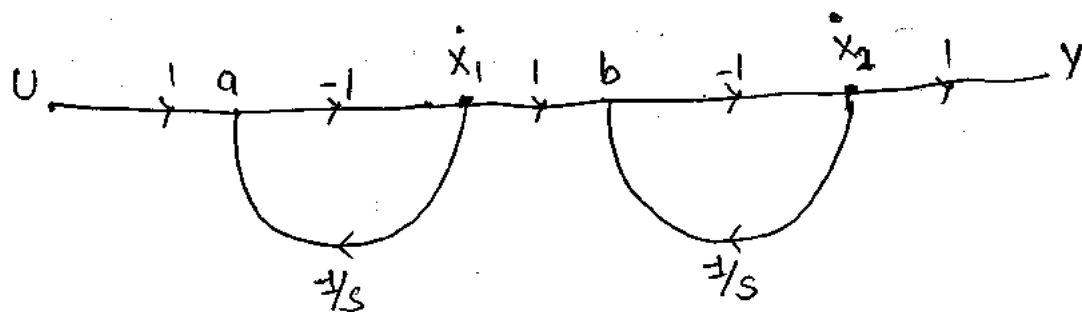
$$b = a + \dot{x}_2$$

$$b = x_1 - \dot{x}_2 + \dot{x}_2$$

$$b = x_1$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 1 & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 2 \\ 0 \end{bmatrix} u$$

$$y = \begin{bmatrix} 0.5 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$



$$a = U + x_1$$

$$b = \dot{x}_1 + x_2$$

Now,

$$\dot{x}_1 = -a = -U - x_1$$

$$\boxed{\dot{x}_1 = -U - x_1}$$

$$Y = \dot{x}_2 = +x_1 - x_2 + U$$

$$\dot{x}_2 = -b$$

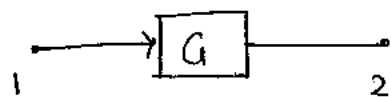
$$\therefore b = -U - x_1 + x_2$$

$$\boxed{\dot{x}_2 = +x_1 - x_2 + U}$$

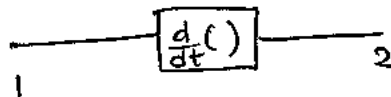
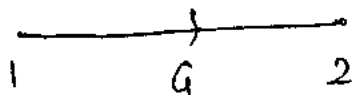
$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \underbrace{\begin{bmatrix} -1 & 0 \\ +1 & -1 \end{bmatrix}}_A \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \underbrace{\begin{bmatrix} -1 \\ 1 \end{bmatrix}}_B U$$

$$Y = \underbrace{[+1 \quad -1]}_C \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \underbrace{[1]}_D U$$

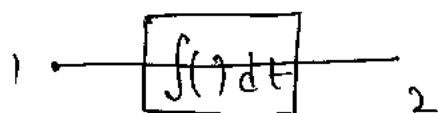
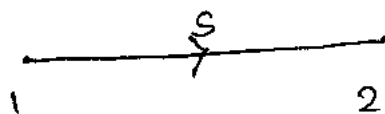
* State model from the block diagram



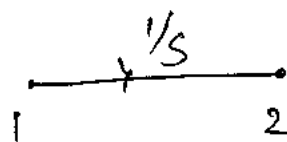
$\Downarrow$



$\Downarrow$

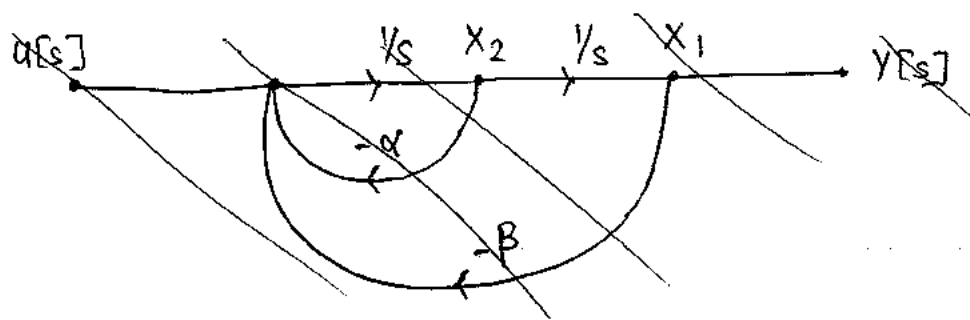
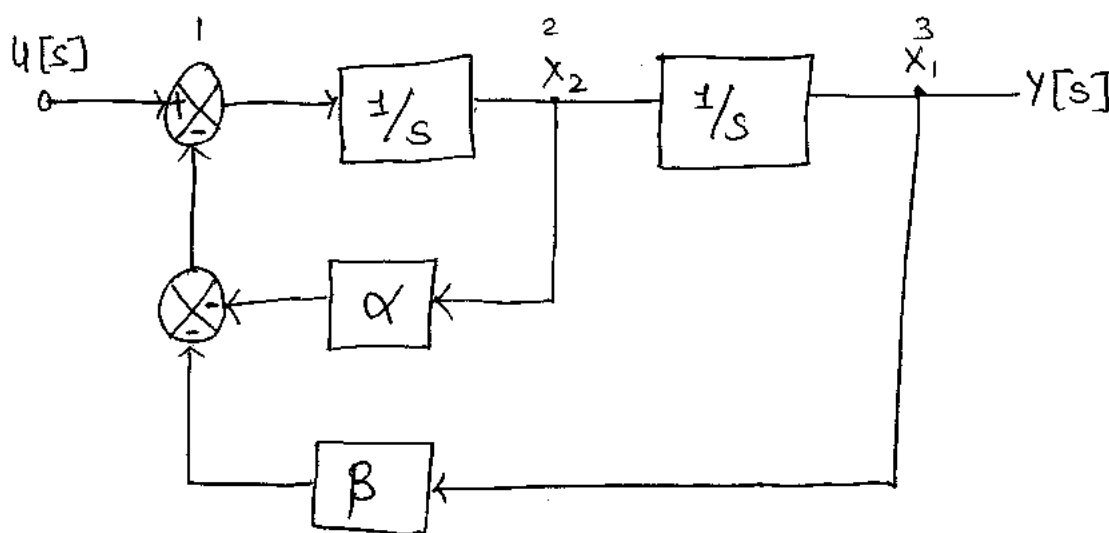


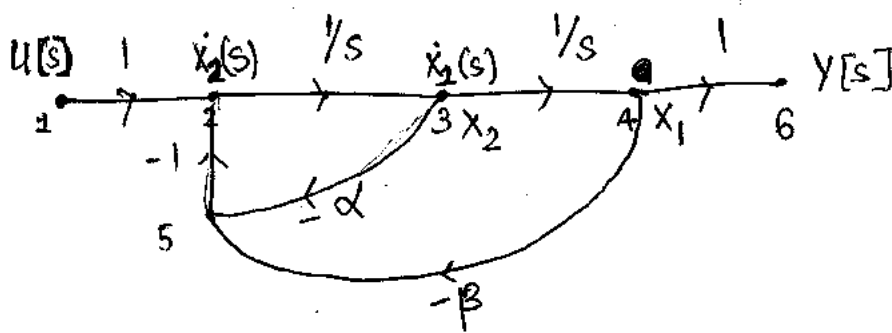
$\Rightarrow$



Block diagram $\rightarrow$ SFG $\rightarrow$ STATE MODEL

Q:- Consider the closed loop system shown in figure
The state model of the system is ____.





$$\dot{x}_2 = \alpha x_2 + \beta x_1 + U$$

$$\dot{x}_1 = x_2$$

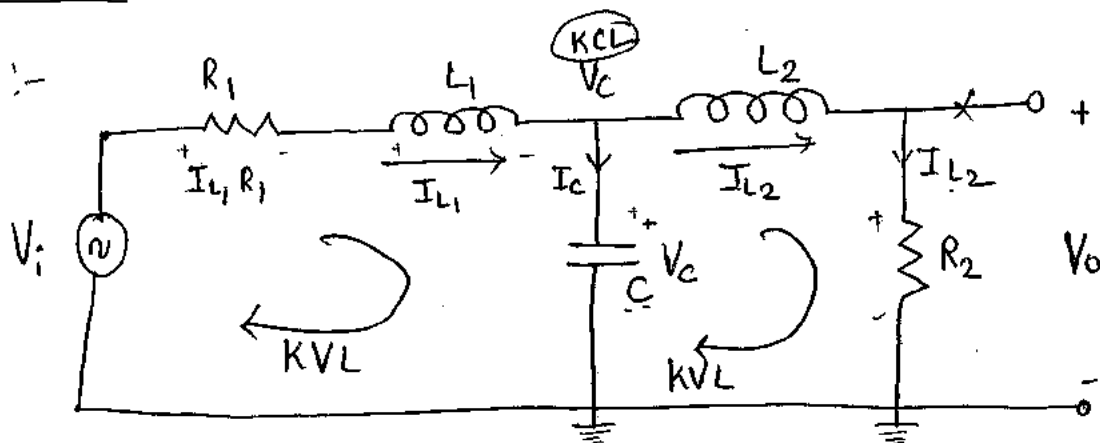
$$Y = x_1$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \beta & \alpha \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} U$$

$$Y = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

*STATE MODEL to ELECTRICAL NETWORK

Q:-



No. of state variable = No. of memory storage elements

No. of " " = 3

1. Select the state variable as current flowing through inductor and voltage across capacitor.
 2. No. of state variables = no. of memory storage element or sum of inductor & capacitor
 3. Write independent KCL, KVL eqⁿ
 4. At capacitance junction apply KCL and apply KVL through inductor
 5. The resultant eqⁿ should consists diff. state variable, state variable, output variable and input variable
- ⇒ No. of state variables = 3

$$\begin{bmatrix} I_{L1} \\ I_{L2} \\ V_C \end{bmatrix}$$

Differential state variables

$$\begin{bmatrix} \dot{I}_{L1} \\ \dot{I}_{L2} \\ \dot{V}_C \end{bmatrix}$$

* KVL₁

$$V_i - I_{L1} R_1 - L_1 \frac{dI_{L1}}{dt} - V_C = 0$$

$$L_1 \dot{I}_{L1} = -I_{L1} R_1 - V_C + V_i$$

$$\boxed{\dot{I}_{L1} = \frac{-R_1}{L_1} I_{L1} - \frac{1}{L_1} V_C + \frac{1}{L_1} V_i} \quad \text{--- (1)}$$

KVL₂

$$V_c - L_2 \frac{dI_{L2}}{dt} - I_{L2} R_2 = 0$$

$$L_2 \frac{dI_{L2}}{dt} = -R_2 I_{L2} + V_c$$

$$L_2 \cdot \dot{I}_{L2} = -R_2 I_{L2} + V_c$$

$$\boxed{\dot{I}_{L2} = \frac{-R_2}{L_2} I_{L2} + \frac{V_c}{L_2}} \quad \text{--- (2)}$$

KCL

$$I_{L1} = I_c + I_{L2}$$

$$I_{L1} = C \frac{dV_c}{dt} + I_{L2}$$

$$C \frac{dV_c}{dt} = I_{L1} - I_{L2}$$

$$C \dot{V}_c = I_{L1} - I_{L2}$$

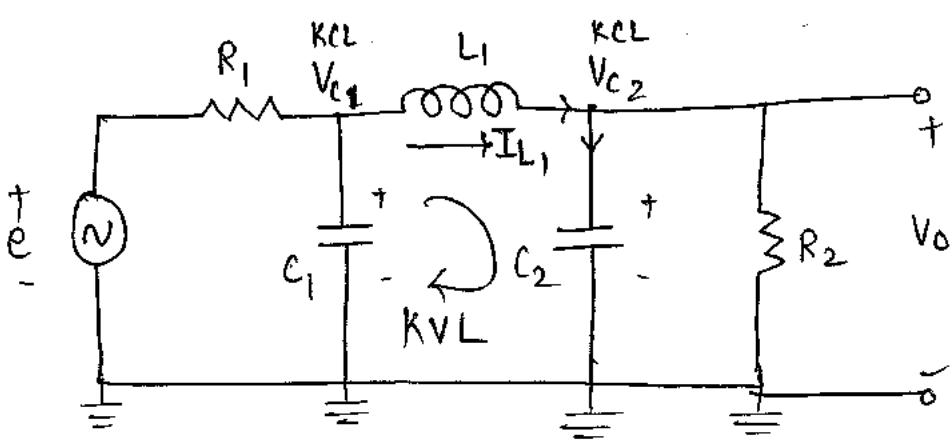
$$\boxed{\dot{V}_c = \frac{I_{L1}}{C} - \frac{I_{L2}}{C}}$$

$$\boxed{V_o = R_2 I_{L2}}$$

$$\begin{bmatrix} \dot{I}_{L1} \\ \dot{I}_{L2} \\ \dot{V}_c \end{bmatrix} = \begin{bmatrix} \frac{-R_1}{L_1} & 0 & -\frac{1}{L_1} \\ 0 & \frac{-R_2}{L_2} & \frac{1}{L_2} \\ \frac{1}{C} & -\frac{1}{C} & 0 \end{bmatrix} \begin{bmatrix} I_{L1} \\ I_{L2} \\ V_c \end{bmatrix} + \begin{bmatrix} \frac{1}{L_1} \\ 0 \\ 0 \end{bmatrix} U$$

$$V_o = [0 \quad R_2 \quad 0] \begin{bmatrix} I_{L1} \\ I_{L2} \\ V_c \end{bmatrix}$$

Q:-



* KCL₁

$$I_{R_1} = I_C + I_{L_1}$$

$$\frac{e}{R_1} = C_1 \frac{dV_{C_1}}{dt} + I_{L_1}$$

$$\frac{e - V_{C_1}}{R_1} = C_1 \dot{V}_{C_1} + I_{L_1}$$

$$\dot{V}_{C_1} = \frac{e}{R_1 C_1} - \frac{I_{L_1}}{C_1}$$

$$\therefore \frac{e}{R_1} - \frac{V_{C_1}}{R_1} = C_1 \dot{V}_{C_1} + I_{L_1}$$

$$\therefore \dot{V}_{C_1} = \frac{e}{R_1 C_1} - \frac{V_{C_1}}{R_1 C_1} - \frac{I_{L_1}}{C_1}$$

* KVL₁

$$L_1 \frac{dI_{L_1}}{dt} + V_{C_2} - V_{C_1} = 0$$

$$L_1 \cdot \dot{I}_{L_1} + V_{C_2} - V_{C_1} = 0$$

$$\dot{I}_{L_1} = \frac{V_{C_1}}{L_1} - \frac{V_{C_2}}{L_2}$$

* KCL₂

$$I_{L_1} = C_2 \frac{dV_{C_2}}{dt} + \frac{V_{C_2} - 0}{R_2}$$

$$I_{L_1} = C_2 \cdot \dot{V}_{C_2} + \frac{V_{C_2}}{R_2}$$

$$\dot{V}_{C_2} = \frac{I_{L_1}}{C_2} - \frac{V_{C_2}}{C_2 R_2}$$

$$\begin{bmatrix} \dot{V}_C \\ \dot{V}_{C2} \\ I_L \end{bmatrix} = \begin{bmatrix} -\frac{1}{R_1 C_1} & 0 & -1/C_1 \\ 0 & -1/C_2 R_2 & 1/C_2 \\ 1/L & -1/L & 0 \end{bmatrix} \begin{bmatrix} V_{C1} \\ V_{C2} \\ I_L \end{bmatrix} + \begin{bmatrix} 1/R_1 C_1 \\ 0 \\ 0 \end{bmatrix} U$$

$$V_0 = \begin{bmatrix} 0 & 1 & 0 \end{bmatrix} \begin{bmatrix} V_{C1} \\ V_{C2} \\ I_L \end{bmatrix}$$

* Transfer Function ~~from~~ State model

$$\dot{X} = AX + BU$$

$$Y = CX + DU$$

$$\text{T.F.} = \frac{L[\text{o/p}]}{L[\text{i/p}]} \Big|_{\text{I.C.}=0}$$

$$= \frac{Y[s]}{U[s]} \Big|_{\text{I.C.}=0}$$

$$sX[s] = A_{n \times n} X[s] + B U[s]$$

$$[sI_{n \times n} - A] X[s] = B U[s]$$

$$\Rightarrow \underline{X[s]} = [sI - A]_{n \times n}^{-1} B_{n \times m} U[s]$$

$$Y[s] = C X[s] + D U[s]$$

$$Y[s] = C_{p \times n} [sI - A]_{n \times n}^{-1} B_{n \times m} U[s] + D \cdot U[s]$$

$$\text{T.F.} = \frac{Y[s]}{U[s]} = C_{p \times n} [sI - A]_{n \times n}^{-1} B_{n \times m} + D_{p \times m}$$

$$\underline{T.F. = C [sI - A]^{-1} B + D}$$

①:-

$$A = \begin{bmatrix} -1 & 0 \\ 1 & -1 \end{bmatrix}, B = \begin{bmatrix} -1 \\ 1 \end{bmatrix}, C = [1 \quad -1], D = 1$$

$$T.F. = C [sI - A]^{-1} B + D$$

$$sI - A = \begin{bmatrix} s & 0 \\ 0 & s \end{bmatrix} - \begin{bmatrix} -1 & 0 \\ 1 & -1 \end{bmatrix} = \begin{bmatrix} s+1 & 0 \\ -1 & s+1 \end{bmatrix}$$

$$\text{adj}^{(sI-A)} = \begin{bmatrix} s+1 & 0 \\ 1 & s+1 \end{bmatrix}$$

$$|sI - A| = (s+1)^2 = s^2 + 2s + 1$$

$$T.F. = [1 \quad -1] \begin{bmatrix} \frac{s+1}{s^2+2s+1} & 0 \\ \frac{1}{s^2+2s+1} & \frac{s+1}{(s+1)^2} \end{bmatrix} \begin{bmatrix} -1 \\ 1 \end{bmatrix} + [1]$$

$$= [1 \quad -1] \begin{bmatrix} \frac{1}{s+1} & 0 \\ \frac{1}{(s+1)^2} & \frac{1}{s+1} \end{bmatrix} \begin{bmatrix} -1 \\ 1 \end{bmatrix} + I$$

$2 \times 2 \quad \quad \quad 2 \times 2$

$$= \begin{bmatrix} \frac{1}{s+1} & -\frac{1}{(s+1)^2} & \frac{1}{s+1} \end{bmatrix} \begin{bmatrix} -1 \\ 1 \end{bmatrix} + I$$

$1 \times 2 \quad \quad \quad 2 \times 1$

$$= \frac{-1}{s+1} + \frac{1}{(s+1)^2} - \frac{1}{s+1} = \frac{-(s+1) + 1 - (s+1)}{(s+1)^2} =$$

$$T.F. = \frac{C \operatorname{adj}[sI-A]B}{|sI-A|} + D$$

$$T.F. = \frac{C \operatorname{adj}[sI-A]B + D|sI-A|}{|sI-A|}$$

Characteristics eqⁿ = $|sI-A|=0 \Rightarrow$ closed loop poles — stable

$$T.F. = \frac{[1 \ -1] \begin{bmatrix} s+1 & 0 \\ 1 & s+1 \end{bmatrix} \begin{bmatrix} -1 \\ 1 \end{bmatrix} + 1|sI-A|}{(s+1)^2}$$

$$= \frac{[1 \ -1] \begin{bmatrix} -(s+1) \\ s \end{bmatrix} + (s+1)^2}{(s+1)^2}$$

$$= \frac{-(s+1) - s + (s+1)^2}{(s+1)^2}$$

$$= \frac{-2s - 1 + s^2 + 2s + 1}{(s+1)^2}$$

$$T.F. = \frac{s^2}{(s+1)^2}$$

Using C.E.

$$T.F. = \frac{1}{1 + \frac{1}{s} + \frac{1}{s} + \frac{1}{s^2}}$$

$$T.F. = \frac{s^2}{s^2 + s + s + 1}$$

$$= \frac{s^2}{(s+1)^2}$$

$$Q:- \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \underset{A}{\begin{bmatrix} -2 & -3 \\ 4 & 2 \end{bmatrix}} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \underset{B}{\begin{bmatrix} 3 \\ 5 \end{bmatrix}} u$$

$$[Y] = \underset{C}{\begin{bmatrix} 1 & 1 \end{bmatrix}} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

Find T.F. of given state model.

$$T.F. = C [sI - A]^{-1} B$$

$$= \begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} s+2 & -3 \\ 4 & s-2 \end{bmatrix}^{-1} \begin{bmatrix} 3 \\ 5 \end{bmatrix}$$

$$= \frac{\begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} s-2 & 3 \\ -4 & s+2 \end{bmatrix} \begin{bmatrix} 3 \\ 5 \end{bmatrix}}{(s+2)(s-2)+12}$$

$$= \frac{\begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} 3s-6+15 \\ -12+5s+10 \end{bmatrix}}{s^2-4+12} = \frac{\begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} 3s+9 \\ 5s-2 \end{bmatrix}}{s^2+8}$$

$$= \frac{3s+9+5s-2}{s^2+8}$$

$$T.F. = \underline{\underline{\frac{8s+7}{s^2+8}}}$$

$$\text{Q:- } \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 3 \\ -2 & -5 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u$$

$$y = \begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

$$\text{Sol:- } \hat{T.F.} = C [sI - A]^{-1} B$$

$$= \begin{bmatrix} 1 & 1 \end{bmatrix} \left[\begin{bmatrix} s & 0 \\ 0 & s \end{bmatrix} - \begin{bmatrix} 0 & 3 \\ -2 & -5 \end{bmatrix} \right]^{-1} \begin{bmatrix} 1 \\ 1 \end{bmatrix}$$

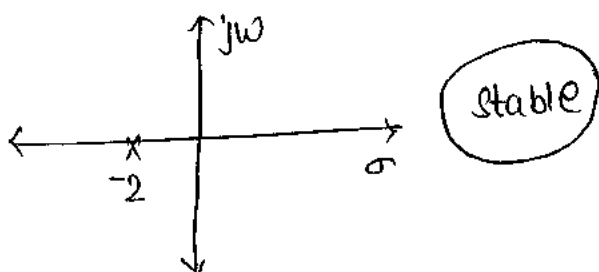
$$= \begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} s & -3 \\ 2 & s+5 \end{bmatrix}^{-1} \begin{bmatrix} 1 \\ 1 \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 1 \end{bmatrix} \frac{\begin{bmatrix} s+5 & 3 \\ -2 & s \end{bmatrix}}{s(s+5)+6} \begin{bmatrix} 1 \\ 1 \end{bmatrix}$$

$$= \frac{\begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} s+8 \\ -2+s \end{bmatrix}}{s(s+5)+6}$$

$$T.F. = \frac{s+8-2+s}{s(s+5)+6} = \frac{2s+6}{s(s+5)+6} = \frac{2s+6}{s^2+5s+6} = \frac{2(s+3)}{(s+3)(s+2)}$$

$$= \frac{2}{s+2}$$



$$\frac{Y[s]}{U[s]} = \frac{2}{s+2}$$

If we want to find I.R.

$$\Rightarrow U[s] = 1$$

$$Y[s] = \frac{2}{s+2}$$

$$y(t) = 2e^{-2t} \cdot u(t)$$

If we want to find USR

$$U[s] = 1/s$$

$$Y[s] = \frac{2}{s(s+2)}$$

$$= 2 \left[\frac{1}{s} - \frac{1}{s+2} \right] \cdot \frac{1}{2}$$

$$y(t) = (1 - e^{-2t})u(t)$$

$$Q:- \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -6 & -5 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$$

$$[Y] = [6 \quad 1] \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

$$T.F. = C(sI - A)^{-1}B$$

$$= [6 \quad 1] \left\{ \begin{bmatrix} s & 0 \\ 0 & s \end{bmatrix} - \begin{bmatrix} 0 & 1 \\ -6 & -5 \end{bmatrix} \right\}^{-1} \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

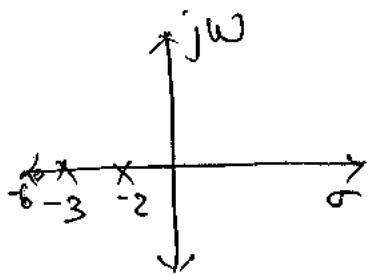
$$= [6 \quad 1] \begin{bmatrix} s & -1 \\ 6 & s+5 \end{bmatrix}^{-1} \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

$$T.F. = [6 \ 1] \frac{\begin{bmatrix} s+5 & 1 \\ -6 & s \end{bmatrix}}{s(s+5)+6} \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

$$= \frac{[6 \ 1] \begin{bmatrix} 1 \\ s \end{bmatrix}}{s(s+5)+6}$$

$$T.F. = \frac{6+s}{s(s+5)+6}$$

$$T.F. = \frac{6+s}{s^2+5s+6} = \frac{s+6}{(s+2)(s+3)}$$



stable

To find system stability,

$$T.F. = \frac{N(s)}{D(s)}$$

$$D(s) = |sI - A| = 0$$

$$\underline{s = -2, -3}$$

(1) I.R.

$$U[s] = 1$$

$$Y[s] = \frac{s+6}{(s+2)(s+3)} = \frac{A}{s+2} + \frac{B}{s+3} = \frac{4}{s+2} + \frac{-3}{s+3}$$

$$y(t) = (4e^{-2t} - 3e^{-3t})u(t)$$

(2) S.R.

$$\frac{Y[s]}{U[s]} = \frac{s+6}{(s+2)(s+3)}$$

$$U[s] = 1/s$$

$$Y[s] = \frac{s+6}{s(s+2)(s+3)}$$

$$= \frac{A}{s} + \frac{B}{s+2} + \frac{C}{s+3}$$

$$= \frac{1}{s} + \frac{-2}{s+2} + \frac{+1}{s+3}$$

$$y(t) = (1 - 2e^{-2t} + e^{-3t}) u(t)$$

* Solution to the state eqⁿs:-

$$\dot{X} = \frac{dx}{dt} = AX + BU \quad \text{--- (1)}$$

$$Y = CX + DU \quad \text{--- (2)}$$

Method: 1 Laplace Transform Method

$$Y[s] = CX[s] + DU[s]$$

$$sX[s] - X(0) = AX[s] + BU[s]$$

$$X[s] = [sI - A]^{-1} X[0] + [sI - A]^{-1} B U[s]$$

$$\downarrow \mathcal{L}^{-1}$$

$$x(t) = \underbrace{\mathcal{L}^{-1} \{ [sI - A]^{-1} X[0] \}}_{\text{ZIR due to input}} + \underbrace{\mathcal{L}^{-1} \{ [sI - A]^{-1} B \cdot U[s] \}}_{\text{ZSR due to input}} \quad \text{--- (1)}$$

ZIR due
to input

ZSR due
to input

Method: 2 Classical Method

$$x(t) = \underbrace{e^{At} x(0)}_{\text{ZIR}} + \int_0^t \underbrace{e^{A(t-\tau)} B u(\tau)}_{\text{ZSR}} \cdot d\tau \quad \text{--- (2)}$$

(1) Compare ZIR term of (1) & (2)

$$\phi(t) = e^{At} = \mathcal{L}^{-1} \{ [sI - A]^{-1} \}$$

↓
State transmission matrix (STM)

$$\phi(s) = [sI - A]^{-1}$$

$$x(t) = e^{At} x(0) + \mathcal{L}^{-1} \{ \phi(s) B \cdot U[s] \}$$

$$y(t) = C [e^{At} x(0) + \mathcal{L}^{-1} \{ \phi(s) \cdot B \cdot U[s] \}] + DU$$

Steps :-

① A, B, C, D, U are given

② $sI - A$

③ $\phi(s) = [sI - A]^{-1}$

④ $\phi(t)$

⑤ $\phi(t) \cdot x(0)$

* Properties of STM :-

$$\Phi(t) = e^{At}$$

$$(1) \Phi(0) = e^0 = I$$

$$\boxed{\Phi(0) = I}$$

$$(2) \Phi^k(t) = e^{Akt} = \Phi(kt)$$

$$\boxed{\Phi^k(t) = \Phi(kt)}$$

$$(3) \boxed{\Phi^{-1}(t) = \Phi(-t)}$$

$$(4) \Phi(t_1 + t_2) = e^{A(t_1 + t_2)} = e^{At_1} \cdot e^{At_2}$$

$$\Phi(t_1 + t_2) = \Phi(t_1) \Phi(t_2)$$

$$\boxed{\Phi(t_1 + t_2) = \Phi(t_1) \Phi(t_2)}$$

$$(5) \boxed{\Phi(t_1 - t_0) \Phi(t_0 - t_2) = \Phi(t_1 - t_2) = \frac{\Phi(t_1)}{\Phi(t_2)}}$$

$$\Phi(t_1 - t_0) \Phi(t_0 - t_2)$$

$$= e^{A(t_1 - t_0)} \cdot e^{A(t_0 - t_2)}$$

$$= e^{A(t_1 - t_2)} = \frac{e^{At_1}}{e^{At_2}} = \frac{\Phi(t_1)}{\Phi(t_2)}$$

$$= \Phi(t_1 - t_2)$$

$$= \frac{\Phi(t_1)}{\Phi(t_2)}$$

Q:- Obtain complete system response of given system below:-

$$\dot{X} = \begin{bmatrix} 0 & 1 \\ -2 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

$$X[0] = \begin{bmatrix} 1 \\ 1 \end{bmatrix}, Y = [1 \quad -1] X$$

Sol: We can't use T.F. method bec. I.C. are given and are not zero.

$$Y = CX + DU$$

$$\text{But } D = 0$$

$$\therefore y(t) = c x(t)$$

$$\Rightarrow x(t) = e^{At} X[0] + \mathcal{L}^{-1} \{ \phi(s) B U(s) \}$$

$$\text{But here } u(t) = 0$$

$$x(t) = e^{At} X[0]$$

$$\phi(t) = e^{At} = [sI - A]^{-1}$$

$$sI - A = \begin{bmatrix} s & -1 \\ 2 & s \end{bmatrix}$$

$$[sI - A]^{-1} = \frac{\text{adj } [sI - A]}{|sI - A|}$$

$$\phi[s] = \frac{\begin{bmatrix} s+1 \\ -2s \end{bmatrix}}{s^2+2} = \begin{bmatrix} \frac{s}{s^2+2} & \frac{+1}{s^2+2} \\ \frac{-2}{s^2+2} & \frac{s}{s^2+2} \end{bmatrix}$$

$$\phi(t) = \begin{bmatrix} \cos\sqrt{2}t & \frac{1}{\sqrt{2}} \sin\sqrt{2}t \\ -\sqrt{2} \sin\sqrt{2}t & \cos\sqrt{2}t \end{bmatrix}$$

$$x(t) = \phi(t) X[0]$$

$$x(t) = \begin{bmatrix} \cos\sqrt{2}t & \frac{1}{\sqrt{2}}\sin\sqrt{2}t \\ -\sqrt{2}\sin\sqrt{2}t & \cos\sqrt{2}t \end{bmatrix} \begin{bmatrix} 1 \\ 1 \end{bmatrix} = \begin{bmatrix} \cos\sqrt{2}t + \frac{1}{\sqrt{2}}\sin\sqrt{2}t \\ -\sqrt{2}\sin\sqrt{2}t + \cos\sqrt{2}t \end{bmatrix}$$

$$y(t) = c x(t)$$

$$= \begin{bmatrix} 1 & -1 \end{bmatrix} \begin{bmatrix} \cos\sqrt{2}t + \frac{1}{\sqrt{2}}\sin\sqrt{2}t \\ -\sqrt{2}\sin\sqrt{2}t + \cos\sqrt{2}t \end{bmatrix}$$

$$y(t) = \cos\sqrt{2}t + \frac{1}{\sqrt{2}}\sin\sqrt{2}t - \sqrt{2}\sin\sqrt{2}t + \cos\sqrt{2}t$$

$$y(t) = \left(\sqrt{2} + \frac{1}{\sqrt{2}}\right) \sin\sqrt{2}t$$

$$\textcircled{Q} \star \dot{x} = \begin{bmatrix} 0 & 1 \\ -2 & 3 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 5 \end{bmatrix} u \quad (\text{IMP})$$

$$x[0] = \begin{bmatrix} 1 \\ 0 \end{bmatrix}, y = [0 \ 1]x. \text{ Find ZSR, ZIR and total response.}$$

$$\underline{\text{Sol}}: Y = CX + DU$$

$$\text{But } D = [0]$$

$$y(t) = c \cdot x(t)$$

$$\Rightarrow x(t) = \underbrace{e^{At} x(0)}_{\text{ZIR}} + \underbrace{L^{-1}\{\phi(s) \cdot B \cdot U(s)\}}_{\text{ZSR}}$$

$$U(s) = 1/s$$

$$[sI - A] = \begin{bmatrix} s & -1 \\ 2 & s-3 \end{bmatrix}$$

$$[sI - A]^{-1} = \frac{\begin{bmatrix} s-3 & 1 \\ -2 & s \end{bmatrix}}{s(s-3)+2}$$

$$\phi[s] = \frac{\begin{bmatrix} s-3 & 1 \\ -2 & s \end{bmatrix}}{s^2-3s+2}$$

$$\phi[s] = \begin{bmatrix} \frac{s-3}{(s-1)(s-2)} & \frac{1}{(s-1)(s-2)} \\ \frac{-2}{(s-1)(s-2)} & \frac{s}{(s-1)(s-2)} \end{bmatrix}$$

$$= \begin{bmatrix} \frac{2}{s-1} + \frac{-1}{s-2} & \frac{-1}{s-1} + \frac{1}{s-2} \\ \frac{2}{s-1} + \frac{-2}{s-2} & \frac{-1}{s-1} + \frac{2}{s-2} \end{bmatrix}$$

$$\phi(t) = \begin{bmatrix} 2e^{+t} - e^{+2t} & -e^{+t} + e^{+2t} \\ 2e^t - 2e^{2t} & -e^t + 2e^{2t} \end{bmatrix}$$

$$\Rightarrow zIR \Rightarrow \phi(t) \cdot x[0] = \begin{bmatrix} 2e^t - e^{2t} & -e^t + e^{2t} \\ 2e^t - 2e^{2t} & -e^t + 2e^{2t} \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$= \begin{bmatrix} 2e^t - e^{2t} \\ 2e^t - 2e^{2t} \end{bmatrix}$$

$$y_{zIR}(t) = C[zIR]$$

$$= [0 \ 1] \begin{bmatrix} 2e^t - e^{2t} \\ 2e^t - 2e^{2t} \end{bmatrix}$$

$$y_{zIR}(t) = 2e^t - 2e^{2t}$$

$$\Rightarrow ZSR \Rightarrow \phi[s] \cdot B \cdot U[s] = \begin{bmatrix} \frac{s-3}{(s-1)(s-2)} & \frac{1}{(s-1)(s-2)} \\ \frac{-2}{(s-1)(s-2)} & \frac{s}{(s-1)(s-2)} \end{bmatrix} \begin{bmatrix} 0 \\ 5 \end{bmatrix} \cdot \frac{1}{s}$$

$$= \begin{bmatrix} \frac{5}{s(s-1)(s-2)} \\ \frac{5s}{s(s-1)(s-2)} \end{bmatrix}$$

$$= \begin{bmatrix} \frac{5}{2s} + \frac{-5}{s-1} + \frac{5}{2(s-2)} \\ \frac{-5}{s-1} + \frac{5}{s-2} \end{bmatrix}$$

$$ZSR \mathcal{L}^{-1}\{\phi(s) \cdot B \cdot U[s]\} = \begin{bmatrix} \frac{5}{2} - 5e^t + \frac{5}{2}e^{2t} \\ -5e^t + 5e^{2t} \end{bmatrix}$$

$$\cancel{ZSR} y_{ZSR}(t) = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} \frac{5}{2} - 5e^t + \frac{5}{2}e^{2t} \\ -5e^t + 5e^{2t} \end{bmatrix}$$

$$y_{ZSR}(t) = -5e^t + 5e^{2t}$$

$$\begin{aligned} \therefore y(t) &= y_{ZIR}(t) + y_{ZSR}(t) \\ &= 2e^t - 2e^{2t} - 5e^t + 5e^{2t} \\ &= (-3e^t + 3e^{2t})u(t) \end{aligned}$$

$$\boxed{y(t) = 3[e^{2t} - e^t]u(t)} \longrightarrow \textcircled{US}$$

Q The state space representation of system is given by
when T.F. of system is _____.

$$\dot{X} = \begin{bmatrix} -1 & 0 \\ 0 & -2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} U$$

$$Y = \begin{bmatrix} 1 \\ 1 \end{bmatrix}^T X$$

As we know,

$$\Rightarrow \text{Here } [D] = 0$$

$$Y = CX + DU$$

$$\therefore y(t) = x(t)$$

$$x(t) = e^{At} x(0) + L^{-1} \{ \Phi(s) B \cdot U(s) \}$$

$$\text{T.F.} = C [sI - A]^{-1} B + D$$

$$= \begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} s+1 & 0 \\ 0 & s+2 \end{bmatrix}^{-1} \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 1 \end{bmatrix} \frac{\begin{bmatrix} s+2 & 0 \\ 0 & s+1 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix}}{(s+1)(s+2)}$$

$$= \begin{bmatrix} 1 & 1 \end{bmatrix} \frac{\begin{bmatrix} s+2 \\ 0 \end{bmatrix}}{(s+1)(s+2)}$$

$$\text{T.F.} = \frac{s+2}{(s+1)(s+2)} = \frac{1}{s+1}$$

Q:- Find T.F. to the given state variable

$$\dot{x}_1 = 2x_1 - x_2 + 3U$$

$$\dot{x}_2 = -4x_2 - U$$

$$y = 3x_1 - 2x_2$$

Ans: T.F. = $C[sI - A]^{-1}B + D$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 2 & -1 \\ 0 & -4 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 3 \\ -1 \end{bmatrix} U$$

$$Y = \begin{bmatrix} 3 & -2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + 0$$

$$A = \begin{bmatrix} 2 & -1 \\ 0 & -4 \end{bmatrix}, B = \begin{bmatrix} 3 \\ -1 \end{bmatrix}, C = \begin{bmatrix} 3 & -2 \end{bmatrix}, D = 0$$

$$\text{T.F.} = \begin{bmatrix} 3 & -2 \end{bmatrix} \begin{bmatrix} s-2 & +1 \\ 0 & +4+s \end{bmatrix}^{-1} \begin{bmatrix} 3 \\ -1 \end{bmatrix} + 0$$

$$= \begin{bmatrix} 3 & -2 \end{bmatrix} \begin{bmatrix} +4+s & -1 \\ 0 & s-2 \end{bmatrix} \begin{bmatrix} 3 \\ -1 \end{bmatrix} \frac{1}{(s-2)(-4-s)}$$

$$= \begin{bmatrix} 3 & -2 \end{bmatrix} \begin{bmatrix} +12+3s+1 \\ -s+2 \end{bmatrix} \frac{1}{(s-2)(+4+s)}$$

$$= \frac{+39+9s+2s-4}{(s-2)(s+4)}$$

$$= \frac{35+11s}{(s-2)(s+4)}$$

Q:- $\dot{x} = \begin{bmatrix} 0 & 1 \\ 0 & -1 \end{bmatrix} x$, $x[0] = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$, $y = [0 \ 1] x$

(A) $\sin t$ Find $y(t)$.

(B) $1 - e^{-t}$

(C) $1 - \cos t$

☒ (D) 0

$$x(t) = e^{At} x[0] + L^{-1} \{ \phi(s) B \cdot U[s] \}$$

$$U[s] = 0$$

$$x(t) = e^{At} x[0]$$

$$\phi(t) = e^{At} = [sI - A]^{-1}$$

$$sI - A = \begin{bmatrix} s & -1 \\ 0 & s+1 \end{bmatrix}$$

$$[sI - A]^{-1} = \frac{\begin{bmatrix} s+1 & 1 \\ 0 & s \end{bmatrix}}{s(s+1)}$$

$$y(t) = \phi(t) \cdot x[t]$$

$$= [0 \ 1] \phi(t) \begin{bmatrix} 1 \\ 0 \end{bmatrix}_{2 \times 1}$$

$$y(t) = 0 \cdot \phi(t) = 0$$

Q:- $A = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$

Find $\phi(t)$ matrix & comment on stability

$$\phi(t) = [sI - A]^{-1}$$

$$= \begin{bmatrix} s-1 & 0 \\ 0 & s-1 \end{bmatrix}^{-1}$$

$$= \begin{bmatrix} s-1 & 0 \\ 0 & s-1 \end{bmatrix}$$

$(s-1)^2 \rightarrow$ Poles location $1, 1$

$$= \begin{bmatrix} \frac{1}{s-1} & 0 \\ 0 & \frac{1}{s-1} \end{bmatrix}$$

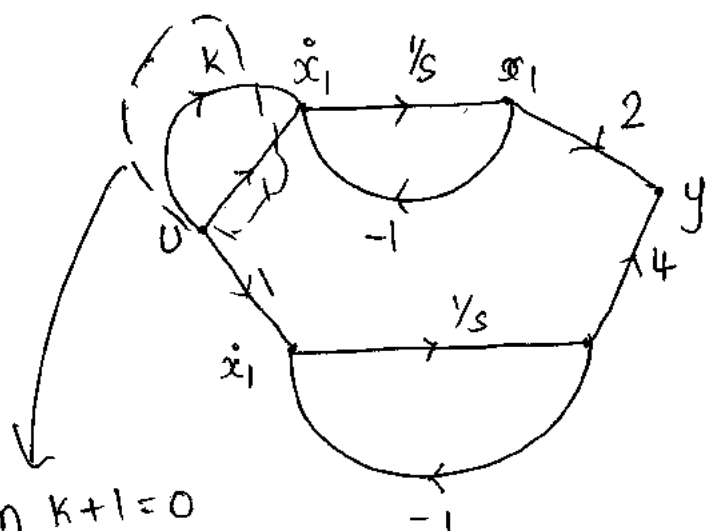
$$\phi(t) = \begin{bmatrix} e^t & 0 \\ 0 & e^t \end{bmatrix} \rightarrow \text{US}$$

* Controllability and Observability

- A system is said to be controllable if it is possible to transfer initial state of system to the desired state of system in a finite time interval, by the controlled input.

- If the SFG is given then to check the controllability observe the continuous path from input to each and every state variable if the path exists then it is controllable.

Q:- Find the k value to become system uncontrollable



Gain $k+1=0$

$k = -1$ → system becomes uncontrollable
 bcz. path doesn't exist to input to state variable.

* Kalman's Test for controllability

Find A, B, C, D matrix

$$Q_c = [B \quad AB \quad A^2B \quad A^3B \quad \dots \quad A^{n-1}B]$$

$$|Q_c| \neq 0 \rightarrow \text{controllable}$$

$$\text{Rank of } Q_c = \text{Rank of } A$$

Q:- Check the controllability of given system:-

$$\frac{Y[s]}{U[s]} = \frac{1}{s^3 + 2s^2 + 3s + 4}$$

$$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -4 & -3 & -2 \end{bmatrix}_{3 \times 3}, \quad B = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}_{3 \times 1}$$

$$AB = \begin{bmatrix} 0 \\ 1 \\ -2 \end{bmatrix}$$

$$A^2B = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -4 & -3 & -2 \end{bmatrix} \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -4 & -3 & -2 \end{bmatrix} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$

$$= \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -4 & -3 & -2 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \\ -2 \end{bmatrix}$$

$$= \begin{bmatrix} 1 \\ -2 \\ -3+4 \end{bmatrix} = \begin{bmatrix} 1 \\ -2 \\ 1 \end{bmatrix}$$

$$A^2B = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -4 & -3 & -2 \end{bmatrix} \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -4 & -3 & -2 \end{bmatrix} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$

$$= \begin{bmatrix} 0 & 0 & 1 \\ -4 & -3 & -2 \\ 8 & 2 & 1 \end{bmatrix} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} \quad \begin{matrix} 4+6 \\ 1 \\ -3+4 \end{matrix}$$

$$A^2B = \begin{bmatrix} 1 \\ -2 \\ 1 \end{bmatrix}$$

$$Q_c = \begin{bmatrix} B & AB & A^2B \\ 0 & 0 & 1 \\ 0 & 1 & -2 \\ 1 & -2 & 1 \end{bmatrix}$$

$$|Q_c| = -1 \neq 0 \rightarrow \text{controllable}$$

* Cont Observability

A system is said to be observable if it is possible to determine the initial state of system by observing the output in finite time interval

* Kalaman's Test for Observability

Find A, B, C, D

$$Q_0 = \begin{bmatrix} C \\ CA \\ CA^2 \\ CA^3 \\ \vdots \\ CA^{n-1} \end{bmatrix}$$

If $|Q_0| \neq 0 \Rightarrow$ observable

Rank of $Q_0 =$ Rank of A $\nearrow$

Q:- Check controllability & observability of system.

$$\dot{x} = \begin{bmatrix} 2 & 0 \\ -1 & 1 \end{bmatrix} x + \begin{bmatrix} 1 \\ -1 \end{bmatrix} u$$

$$y = \begin{bmatrix} 1 & 1 \end{bmatrix} x$$

$$A = \begin{bmatrix} 2 & 0 \\ -1 & 1 \end{bmatrix}, \quad B = \begin{bmatrix} 1 \\ -1 \end{bmatrix}$$

$$AB = \begin{bmatrix} 2 \\ 1-2 \end{bmatrix}$$

$$A^2 B = \begin{bmatrix} 2 & 0 \\ -1 & 1 \end{bmatrix} \begin{bmatrix} 2 \\ 0 \end{bmatrix} = \begin{bmatrix} 4 \\ -2 \end{bmatrix}$$

$$Q_c = \begin{bmatrix} 1 & 2 \\ -1 & -2 \end{bmatrix}$$

$$|Q_c| = -2 + 2 = 0$$

$|Q_c| = 0 \Rightarrow$ not controllable

$$CA = \begin{bmatrix} 1 & 1 \end{bmatrix}_{1 \times 2} \begin{bmatrix} 2 & 0 \\ -1 & 1 \end{bmatrix}_{2 \times 2}$$

$$CA = \begin{bmatrix} 1 & 1 \end{bmatrix}$$

$$Q_o = \begin{bmatrix} 1 & 1 \\ 1 & 1 \end{bmatrix} = 0$$

$|Q_o| = 0 \Rightarrow$ unobservable

$$Q:- \dot{x}_1 = -2x_1 + x_2 + u$$

$$\dot{x}_2 = -x_2 + u$$

$$y = x_1 + x_2$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -2 & 1 \\ 0 & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u$$

$$y = \begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + 0$$

$$A = \begin{bmatrix} -2 & 1 \\ 0 & -1 \end{bmatrix}, B = \begin{bmatrix} 1 \\ 1 \end{bmatrix}, AB = \begin{bmatrix} -1 \\ -1 \end{bmatrix}$$

$$Q_c = \begin{bmatrix} 1 & -1 \\ 1 & -1 \end{bmatrix} = -1 + 1 = 0 \Rightarrow \text{uncontrollable}$$

$$C = \begin{bmatrix} 1 & 1 \end{bmatrix}$$

$$CA = \begin{bmatrix} 1 & 1 \end{bmatrix}_{1 \times 2} \begin{bmatrix} -2 & 1 \\ 0 & -1 \end{bmatrix}_{2 \times 2}$$

$$= \begin{bmatrix} -2 & 0 \end{bmatrix}$$

$$Q_0 = \begin{bmatrix} 1 & 1 \\ -2 & 0 \end{bmatrix} = 0 + 2 = 2 \neq 0 \Rightarrow \text{Observable}$$